

What theorists should know when working with experimentalists

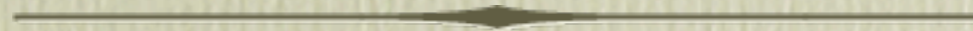
Marcos Curty
University of Vigo

QCrypt 2013

OUTLINE

- Motivation
- Characterisation of experimental components
- QKD with decoy states (asymptotic case)
- Parameter estimation (finite case)
- Side-channels

Motivation



MOTIVATION


















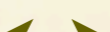
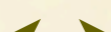


From a theoretical point of view, a QKD system is rather simple. For instance, in the BB84 protocol:

Signals sent by Alice:

Bob's measurements:

Bob's results:

Sifted bits:

						
						
						
1	×	0	×	0	1	×

C.H. Bennett and G. Brassard, Proc. IEEE International Conference on Computers, Systems, and Signal Processing, Bangalore, India, (IEEE, New York), p. 175 (1984).

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Secret key rate:

$$K \propto 1 - h(e_{\text{bit}}) - h(e_{\text{phase}})$$

P.W. Shor and J. Preskill, PRL 85, 441 (2000).

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In practice, however, the situation looks less simple.



QPN 5505 commercial QKD system from MagiQ Technologies (Image taken from <http://www.vad1.com>)

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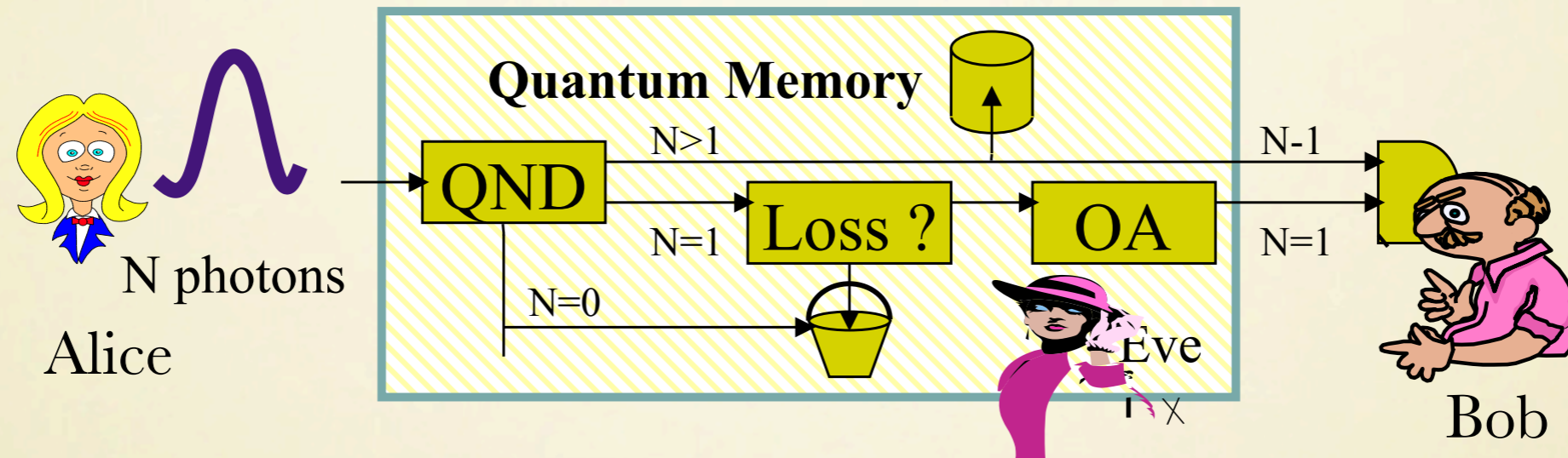
- Alice can emit signals that contain more than one photon prepared in the same polarisation state.
- Bob's detectors can output a double ``click'' due, for example, to dark counts.



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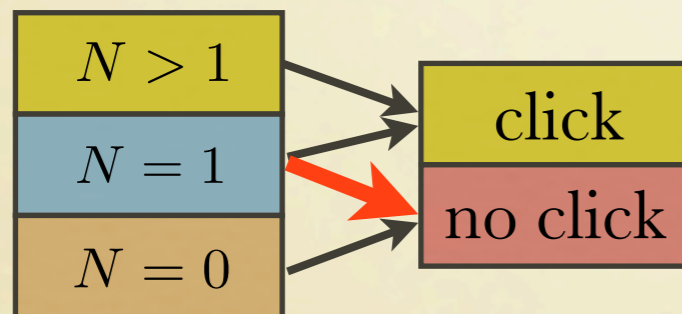
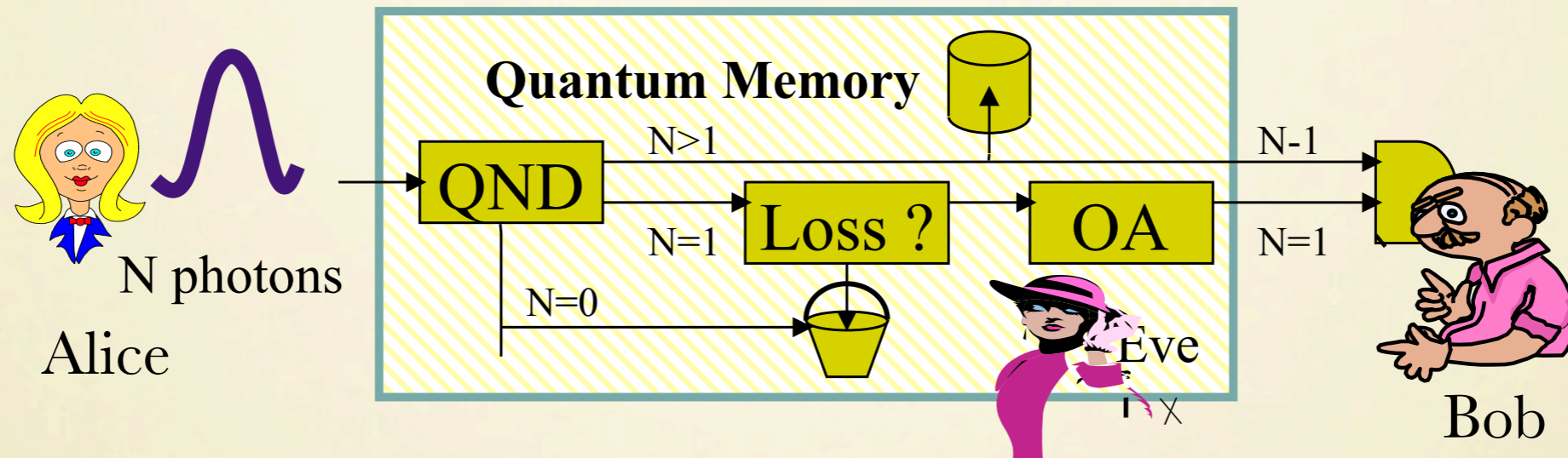
Example: Photon number splitting (PNS) attack.



B. Huttner et al., PRA 51, 1863 (1995); G. Brassard et al., PRL 85, 1330 (2000).

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Eve has full information about the part of the key generated from multi-photon signals

$$K \leq p_{\text{exp}} - p_{\text{multi}}$$

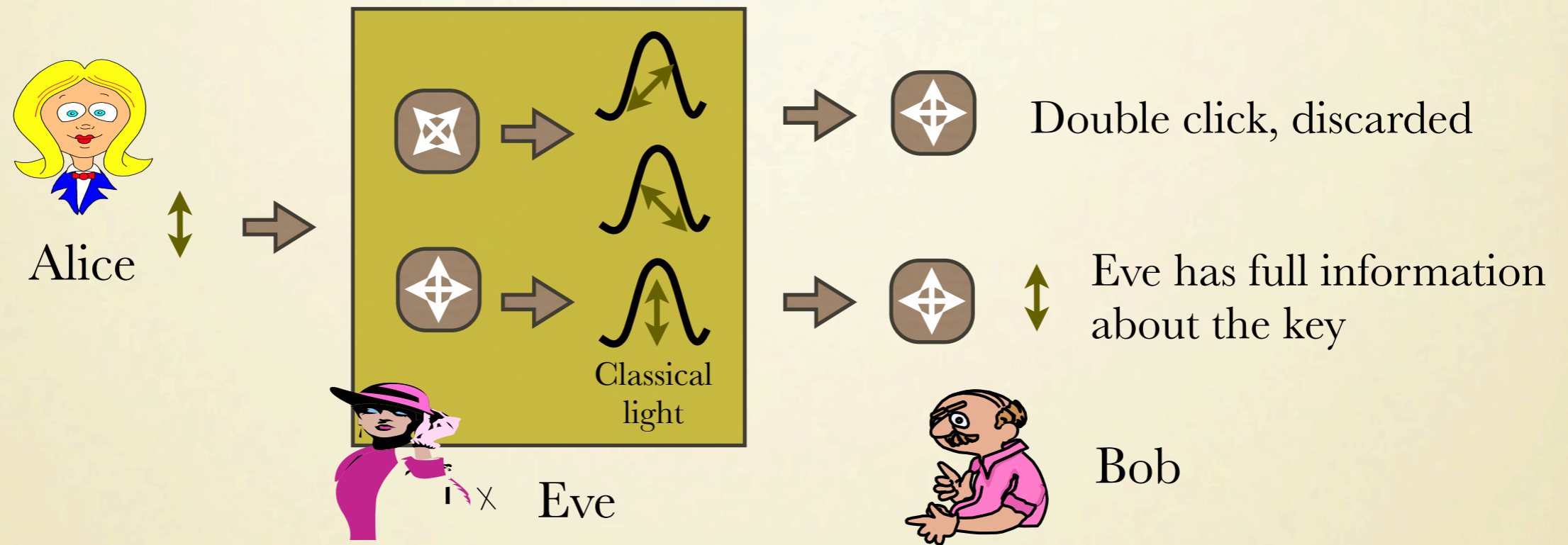
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Example: Exploiting double-clicks (if Bob discards them).

MOTIVATION

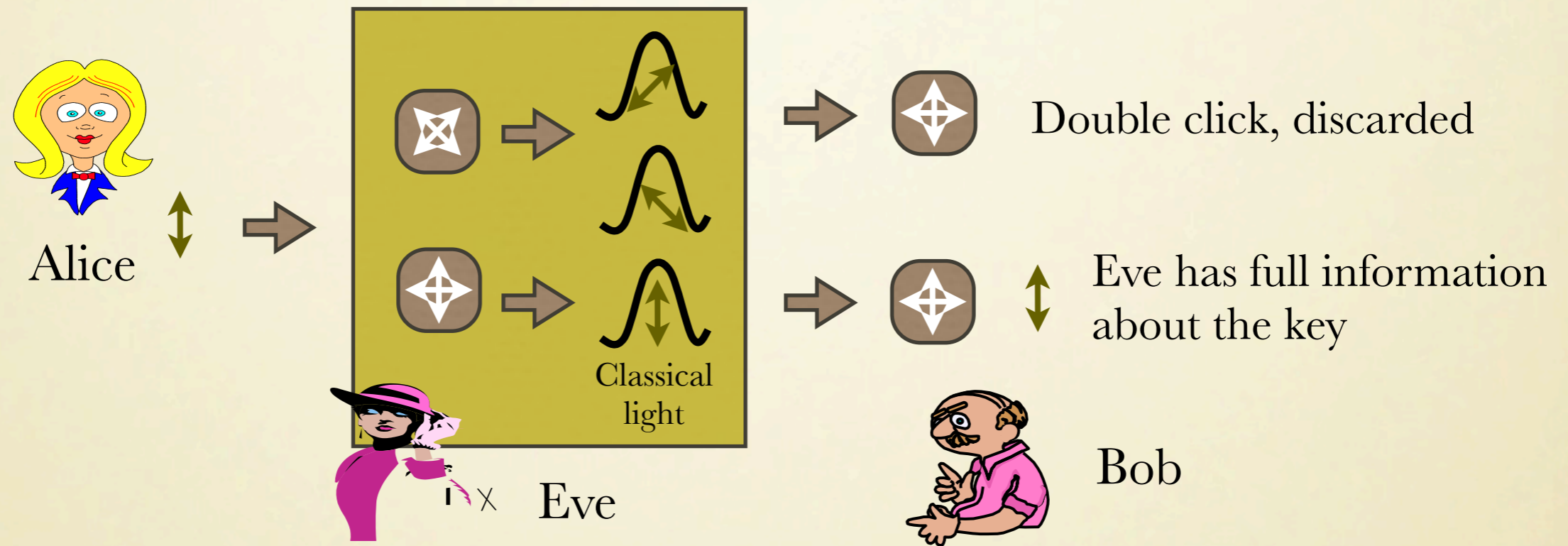
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There is a gap between theory and practice. Theorists have to develop security proofs that can be applied to the experimental realisations.

Characterisation of experimental components



CHARACTERISATION OF PRACTICAL DEVICES

Phase-randomised weak coherent pulses:

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Coherent states: $|\alpha e^{i\phi}\rangle = e^{-|\alpha|^2/2} \sum_{n=0}^{\infty} \frac{(\alpha e^{i\phi})^n}{\sqrt{n!}} |n\rangle$

$$|n\rangle = \frac{1}{\sqrt{n!}} (a^\dagger)^n |0\rangle$$



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If the phase is randomised, we have:

$$\rho = \frac{1}{2\pi} \int_{\phi} |\alpha e^{i\phi}\rangle \langle \alpha e^{i\phi}| d\phi = e^{-|\alpha|^2} \sum_{n=0}^{\infty} \frac{|\alpha|^{2n}}{n!} |n\rangle \langle n|$$

$\mu = |\alpha|^2$
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 $= e^{-\mu} \sum_{n=0}^{\infty} \frac{\mu^n}{n!} |n\rangle \langle n|$

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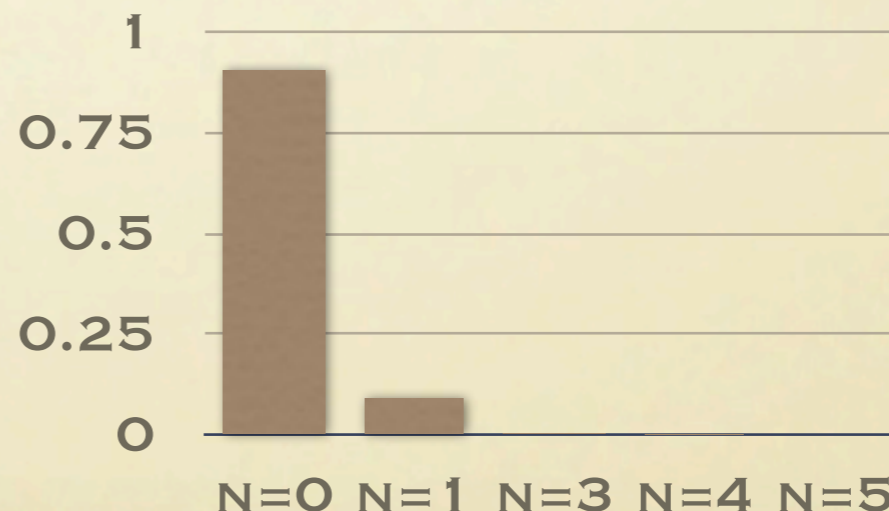


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Photon number statistics
when the intensity $\mu = 0.1$



CHARACTERISATION OF PRACTICAL DEVICES

The BB84 signals can then be described as:

$$\rho_i = e^{-\mu} \sum_{n=0}^{\infty} \frac{\mu^n}{n!} |n_i\rangle \langle n_i| \quad \text{with} \quad |n_i\rangle = \frac{1}{\sqrt{n!}} (a_i^\dagger)^n |0\rangle$$

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The creation operators a_i can be expressed as a function of two creation operators b_1, b_2 associated to orthogonal polarisations:

creation operators

$$a_H^\dagger = \frac{1}{\sqrt{2}} (b_1^\dagger + b_2^\dagger)$$

$$a_V^\dagger = \frac{1}{\sqrt{2}} (b_1^\dagger - b_2^\dagger)$$

$$a_{+45^\circ}^\dagger = \frac{1}{\sqrt{2}} (b_1^\dagger + ib_2^\dagger)$$

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single photon components

$$|1_H\rangle = a_H^\dagger |0\rangle = \frac{1}{\sqrt{2}} (|1, 0\rangle + |0, 1\rangle)$$

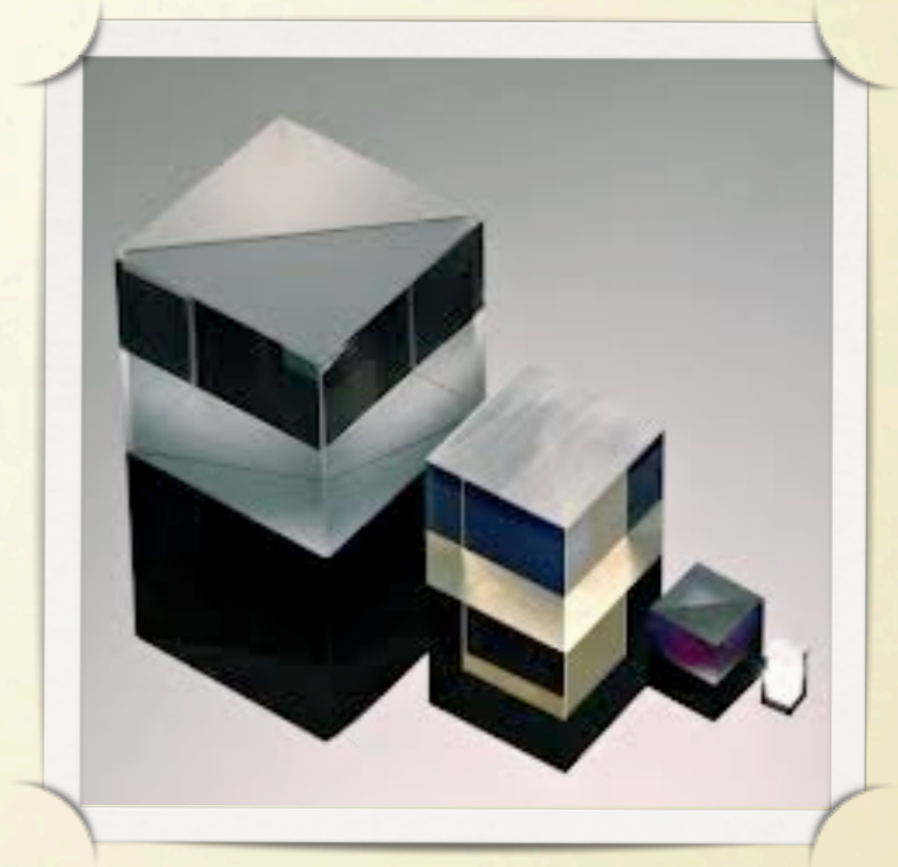
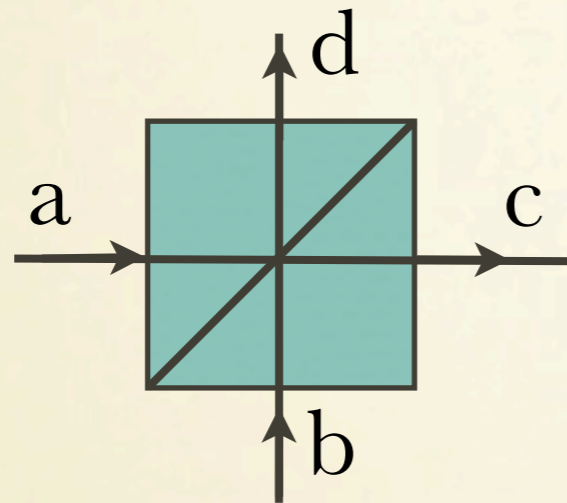
$$|1_V\rangle = a_V^\dagger |0\rangle = \frac{1}{\sqrt{2}} (|1, 0\rangle - |0, 1\rangle)$$

$$|1_{+45^\circ}\rangle = a_{+45^\circ}^\dagger |0\rangle = \frac{1}{\sqrt{2}} (|1, 0\rangle + i|0, 1\rangle)$$

$$|1_{-45^\circ}\rangle = a_{-45^\circ}^\dagger |0\rangle = \frac{1}{\sqrt{2}} (|1, 0\rangle - i|0, 1\rangle)$$

CHARACTERISATION OF PRACTICAL DEVICES

Beam-splitters (BS):



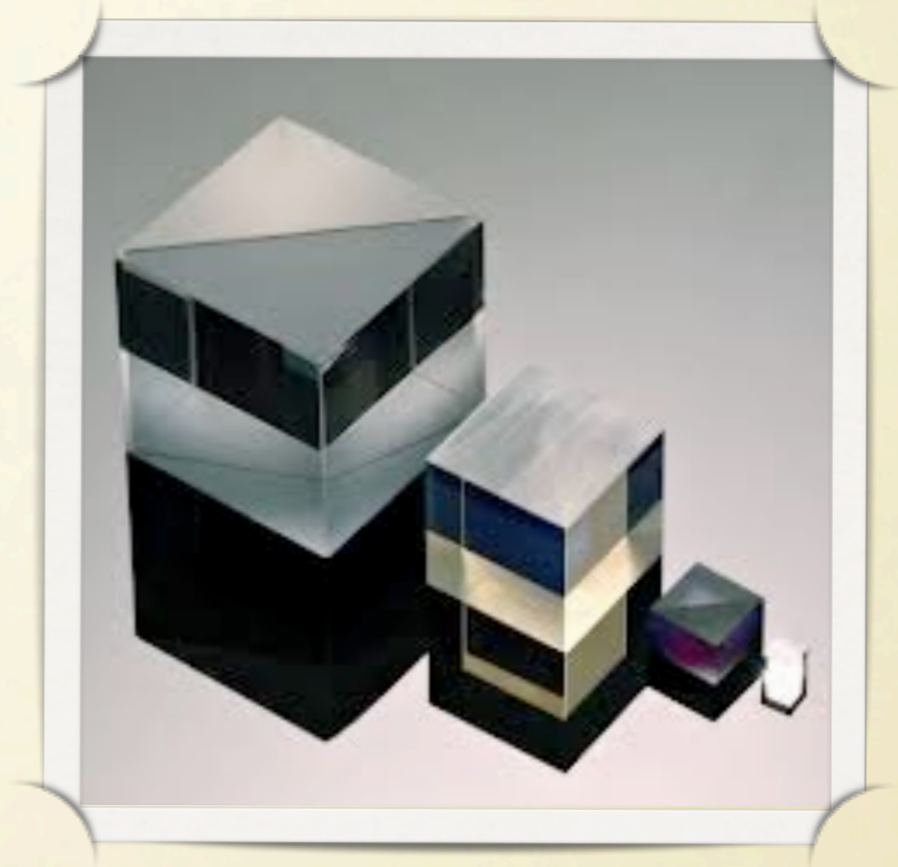
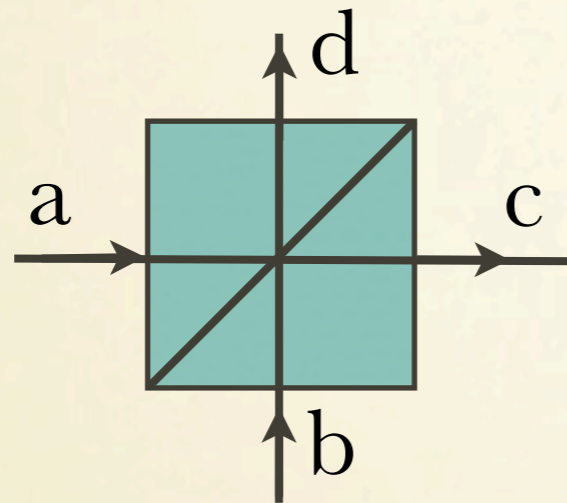
There are two input modes and two output modes

If we neglect for the moment absorption and other imperfections:

$$\begin{pmatrix} a^\dagger \\ b^\dagger \end{pmatrix} = e^{i\phi} \begin{pmatrix} te^{i\phi_t} & re^{i\phi_r} \\ -re^{-i\phi_r} & te^{-i\phi_t} \end{pmatrix} \begin{pmatrix} c^\dagger \\ d^\dagger \end{pmatrix}$$

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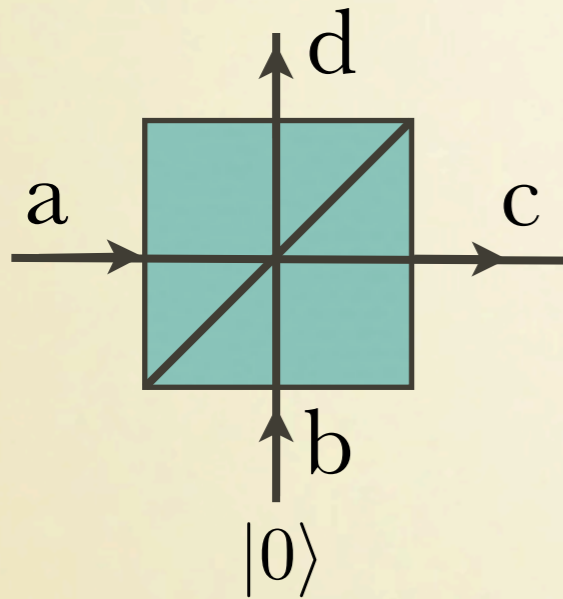
$$50:50 \text{ BS} \rightarrow \begin{pmatrix} a^\dagger \\ b^\dagger \end{pmatrix} = \frac{1}{\sqrt{2}} \begin{pmatrix} 1 & 1 \\ -1 & 1 \end{pmatrix} \begin{pmatrix} c^\dagger \\ d^\dagger \end{pmatrix}$$

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Modelling the losses in the quantum channel (beam-splitter):

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$$\begin{pmatrix} a^\dagger \\ b^\dagger \end{pmatrix} = \begin{pmatrix} \sqrt{\eta_{\text{channel}}} & \sqrt{1 - \eta_{\text{channel}}} \\ -\sqrt{1 - \eta_{\text{channel}}} & \sqrt{\eta_{\text{channel}}} \end{pmatrix} \begin{pmatrix} c^\dagger \\ d^\dagger \end{pmatrix}$$

$$\begin{cases} a^\dagger = \sqrt{\eta_{\text{channel}}}c^\dagger + \sqrt{1 - \eta_{\text{channel}}}d^\dagger \\ b^\dagger = -\sqrt{1 - \eta_{\text{channel}}}c^\dagger + \sqrt{\eta_{\text{channel}}}d^\dagger \end{cases}$$

where $\eta_{\text{channel}} = 10^{-\frac{\alpha d}{10}}$, with:

α represents the loss coefficient of the channel measured in dB/km (e.g. in an optical fibre $\alpha = 0.2$ dB/km)

d is the transmission distance measured in km.

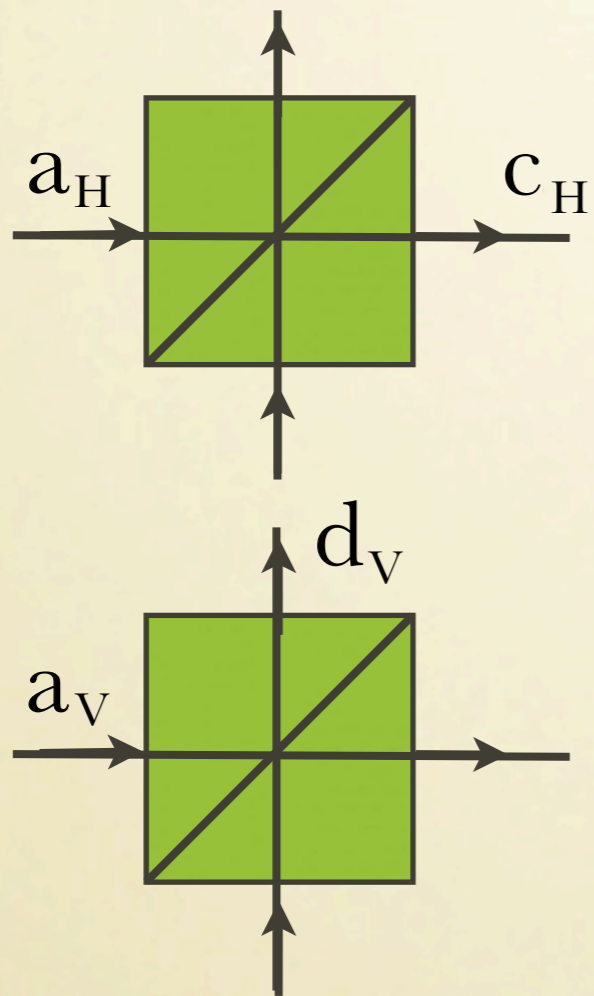
CHARACTERISATION OF PRACTICAL DEVICES

Polarised beam-splitters (PBS):

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Polarised beam-splitters (PBS):

Separate polarisation into spatial modes



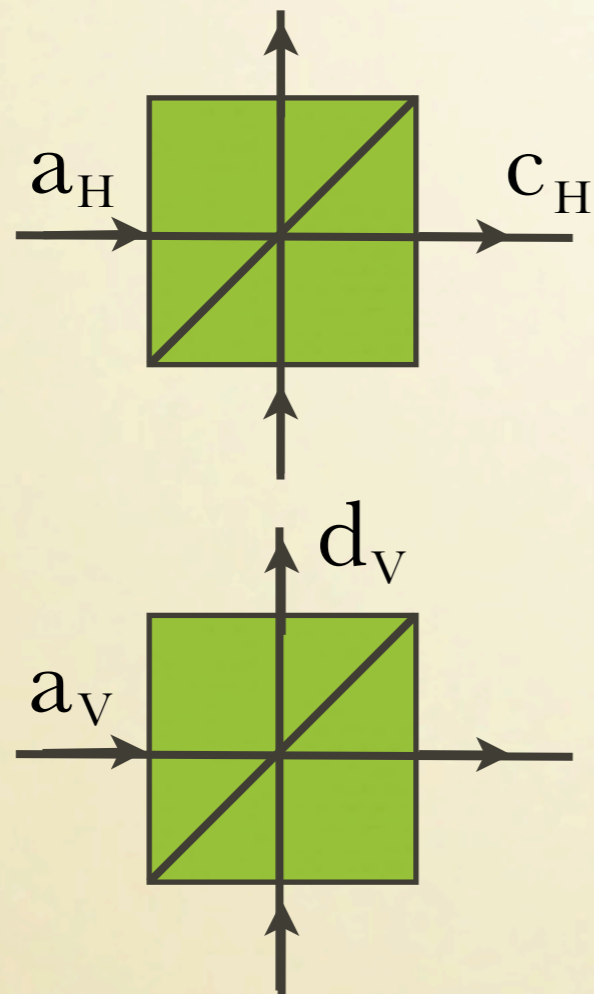
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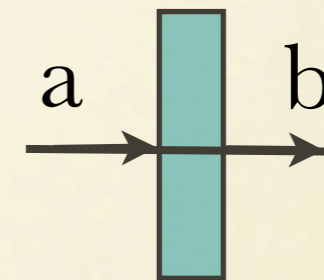


$$a_H^\dagger = c_H^\dagger$$

$$a_V^\dagger = d_V^\dagger$$

Half wave plate (HWP):

Performs a polarisation transformation



$$\begin{pmatrix} a_{+45^\circ}^\dagger \\ a_{-45^\circ}^\dagger \end{pmatrix} = \frac{e^{-i\pi/4}}{\sqrt{2}} \begin{pmatrix} 1 & i \\ 1 & -i \end{pmatrix} \begin{pmatrix} b_{+45^\circ}^\dagger \\ b_{-45^\circ}^\dagger \end{pmatrix}$$

$$a_{+45^\circ}^\dagger = b_V^\dagger$$

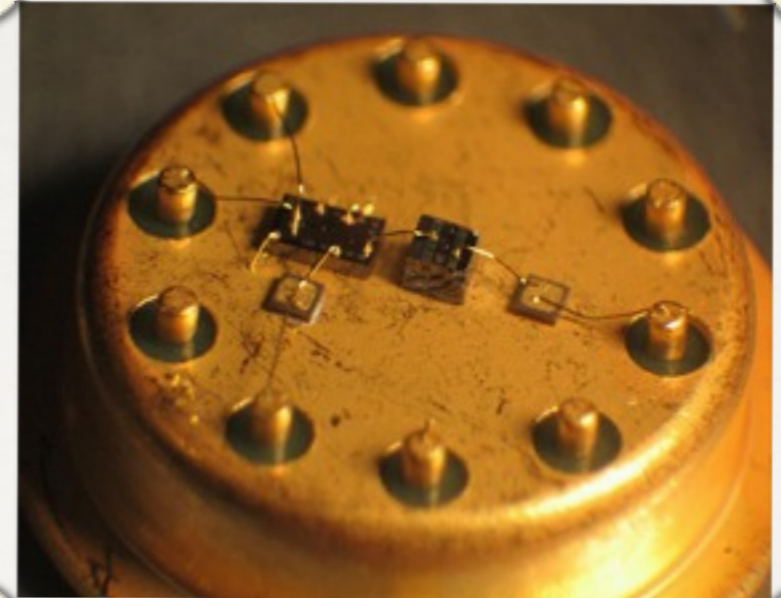
$$a_{-45^\circ}^\dagger = -ib_H^\dagger$$

CHARACTERISATION OF PRACTICAL DEVICES

Threshold detectors:

They provide only two possible outcomes:

- “**Click**”: At least one photon is detected
- “**No click**”: No photon is detected



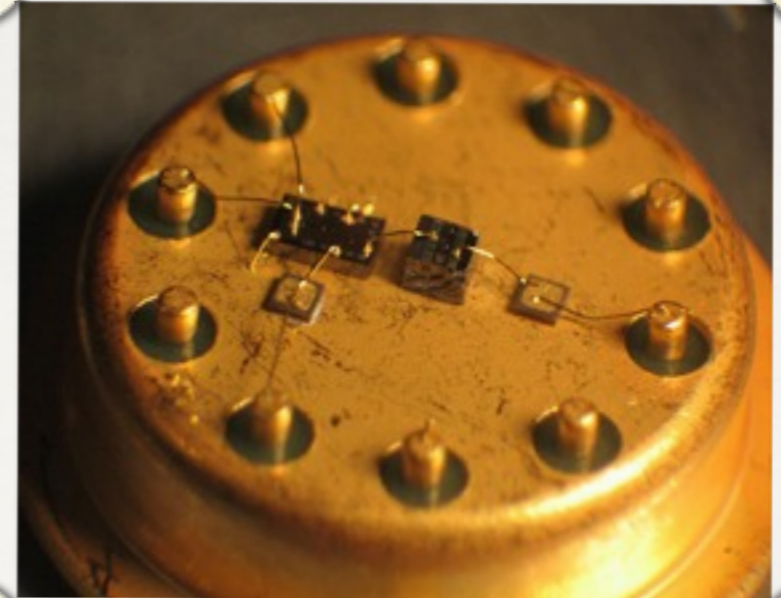
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For simplicity, if we only consider their detection efficiency and dark count rate



$$D_{\text{noclick}} = (1 - p_{\text{dark}}) \sum_{n=0}^{\infty} (1 - \eta_{\text{det}})^n |n\rangle \langle n|$$
$$D_{\text{click}} = 1 - D_{\text{noclick}}$$

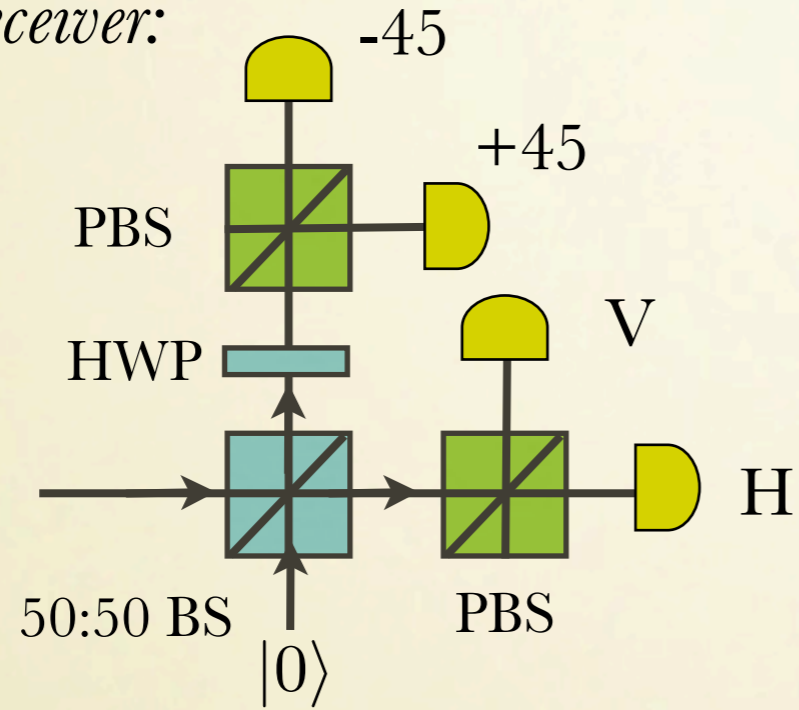
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Example: BB84 receiver.

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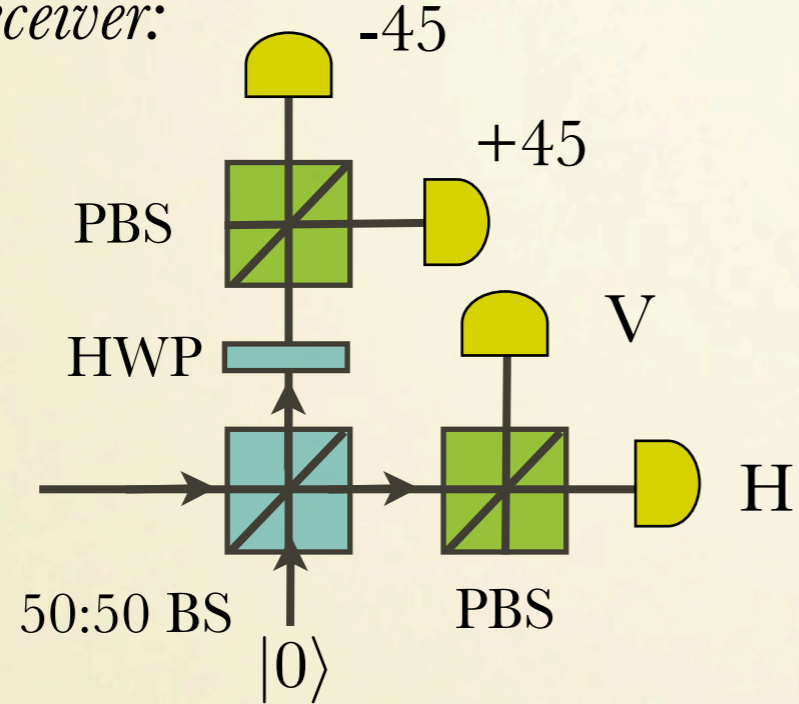
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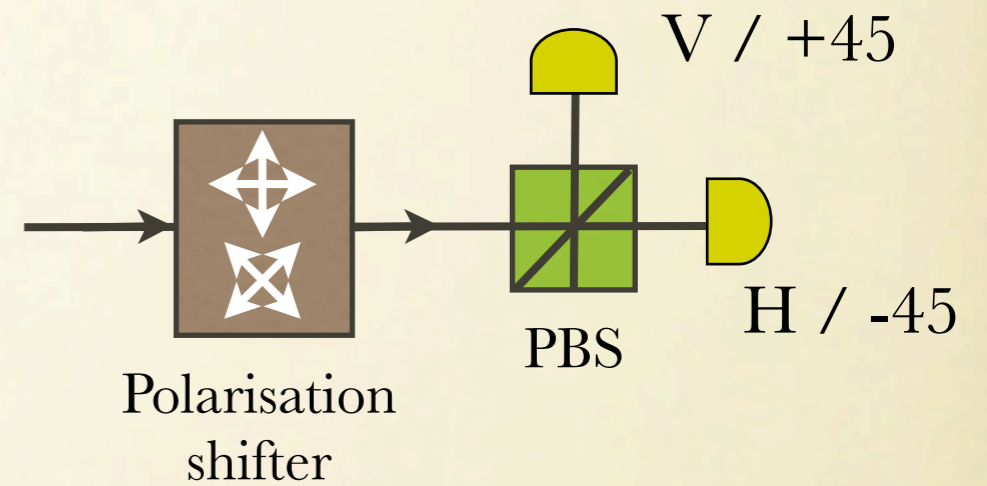
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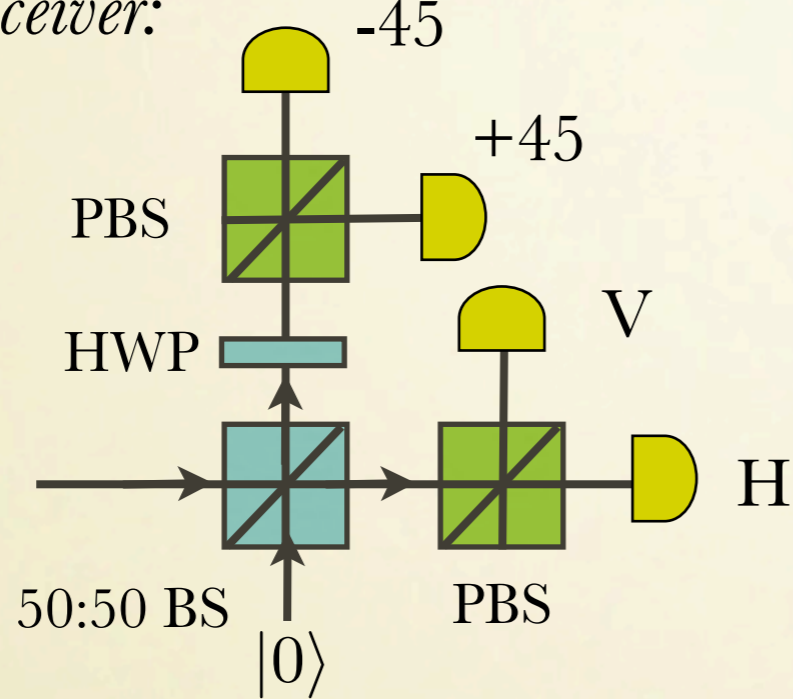
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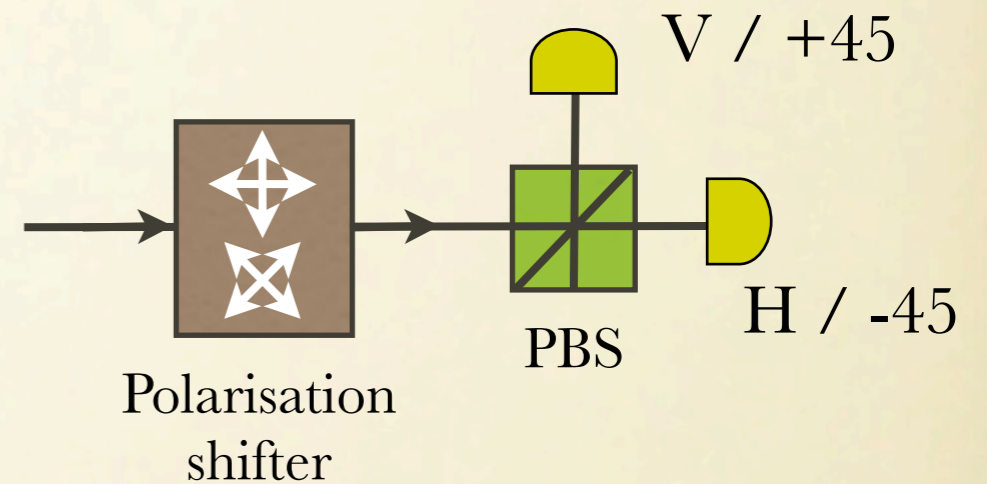
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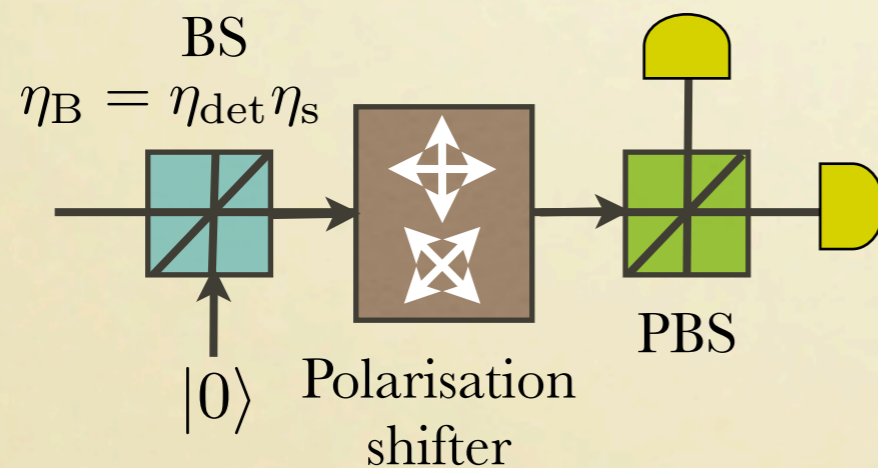
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Active receiver:



If we consider, for the moment, that all detectors have the same efficiency:



$$D_{\text{noclick}} = (1 - p_{\text{dark}}) |0\rangle\langle 0|$$

$$D_{\text{click}} = 1 - D_{\text{noclick}}$$

η_B : Transmittance of the optical components within Bob's measurement device and the detector efficiency

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Example: Gain of a signal state

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The gain Q is defined as the probability that a signal state sent by Alice produces at least one “click” in Bob’s detection apparatus

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The **yield** Y_n of an n -photon state is the conditional probability of a detection event on Bob’s side given that Alice sent an n -photon state

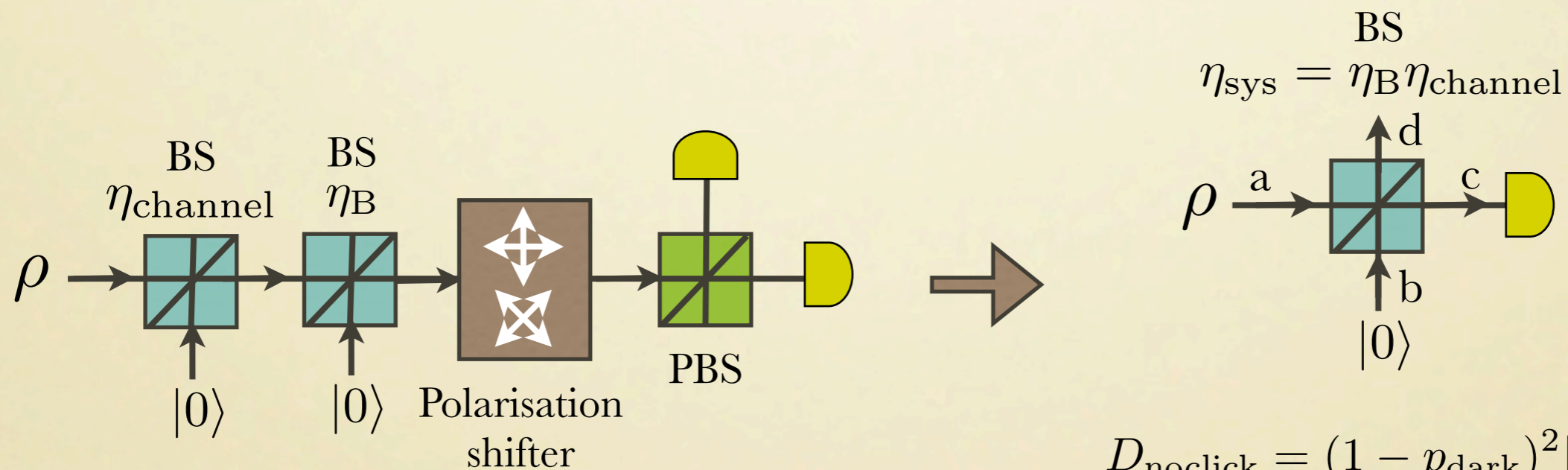
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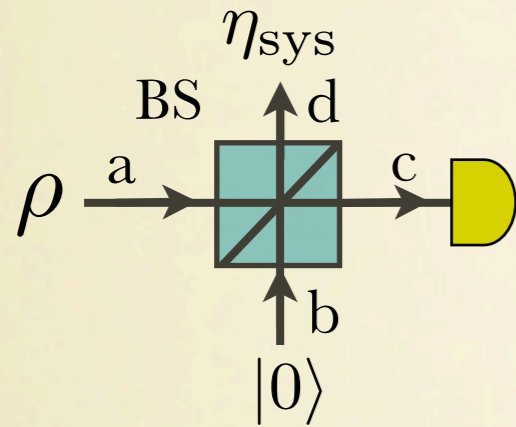
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CHARACTERISATION OF PRACTICAL DEVICES



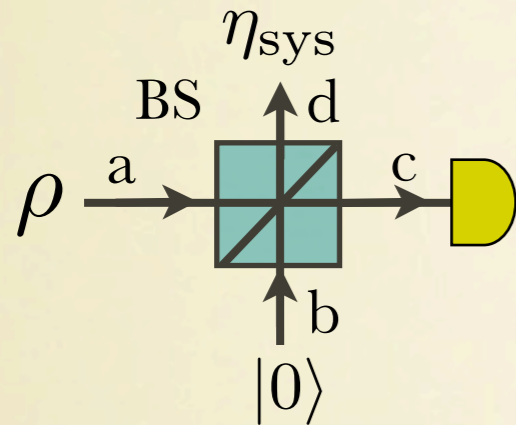
$$|n\rangle_a = \frac{1}{\sqrt{n!}} (a^\dagger)^n |0\rangle \quad \xrightarrow{\text{BS}} \quad |n\rangle_{cd} = \frac{1}{\sqrt{n!}} (\sqrt{\eta_{\text{sys}}} c^\dagger + \sqrt{1 - \eta_{\text{sys}}} d^\dagger)^n |0\rangle$$

$$|n\rangle_{cd} = \sum_{k=0}^n \sqrt{\binom{n}{k}} \sqrt{\eta_{\text{sys}}^{n-k}} \sqrt{1 - \eta_{\text{sys}}}^k |n - k, k\rangle_{cd}$$

Here we have used the fact that

$$|n - k\rangle_c = \frac{1}{\sqrt{(n - k)!}} (c^\dagger)^{n-k} |0\rangle \quad \text{and} \quad |k\rangle_d = \frac{1}{\sqrt{k!}} (d^\dagger)^k |0\rangle$$

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$$\begin{aligned} Y_n &= \text{Tr}[|n\rangle_{cd} \langle n| (D_{\text{click}} \otimes 1_d)] \\ &= 1 - \text{Tr}[|n\rangle_{cd} \langle n| (D_{\text{noclick}} \otimes 1_d)] \\ &= 1 - (1 - p_{\text{dark}})^2 \text{Tr}[|n\rangle_{cd} \langle n| (|0\rangle_c \langle 0| \otimes 1_d)] \\ &= 1 - (1 - p_{\text{dark}})^2 (1 - \eta_{\text{sys}})^n \end{aligned}$$

$$\langle n|m\rangle = \delta_{nm} \quad \rightarrow$$

CHARACTERISATION OF PRACTICAL DEVICES

Given that: $Y_n = 1 - (1 - p_{\text{dark}})^2(1 - \eta_{\text{sys}})^n$

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The gain is directly observed in the experiment.

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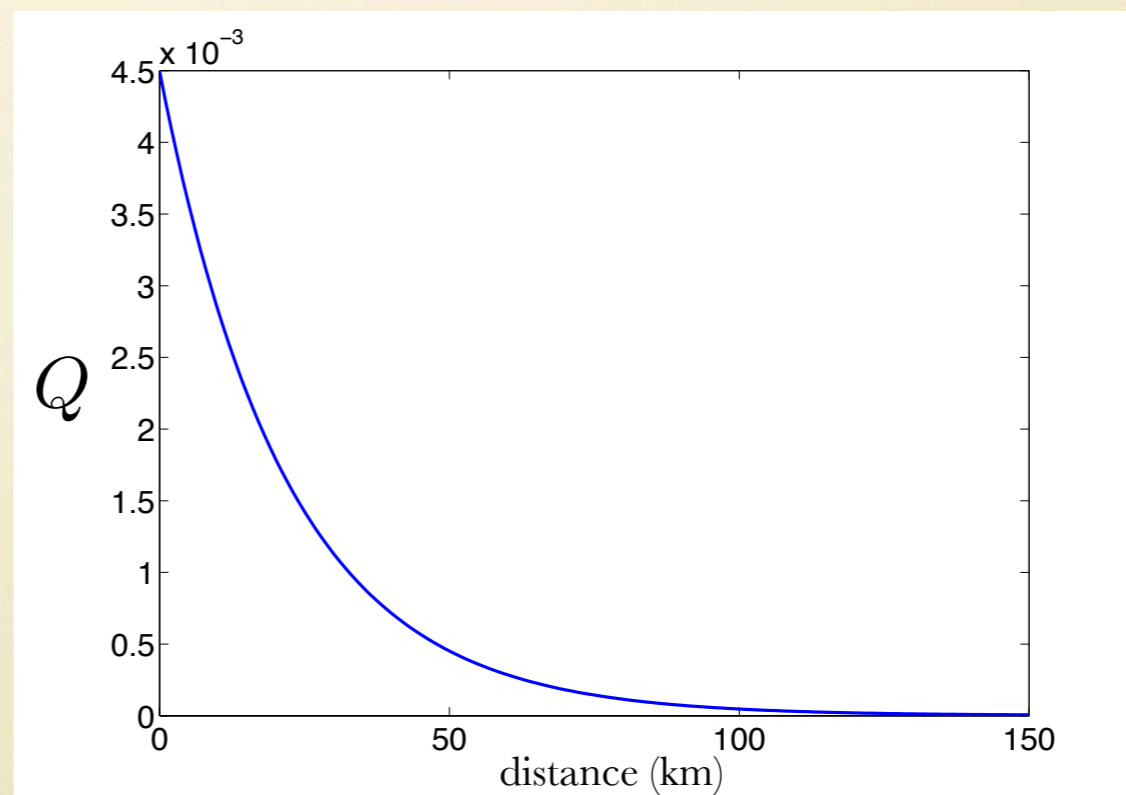
Example:

$$p_{\text{dark}} = 10^{-6}$$

$$\mu = 0.1$$

$$\eta_B = 0.045$$

$$\alpha = 0.2 \text{ dB/km}$$

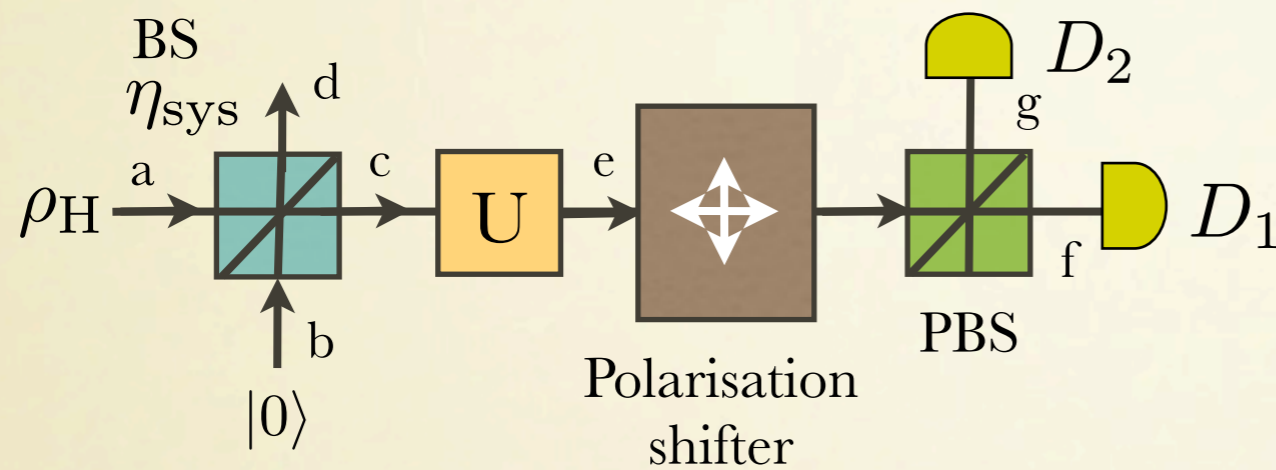


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Example: Error rate

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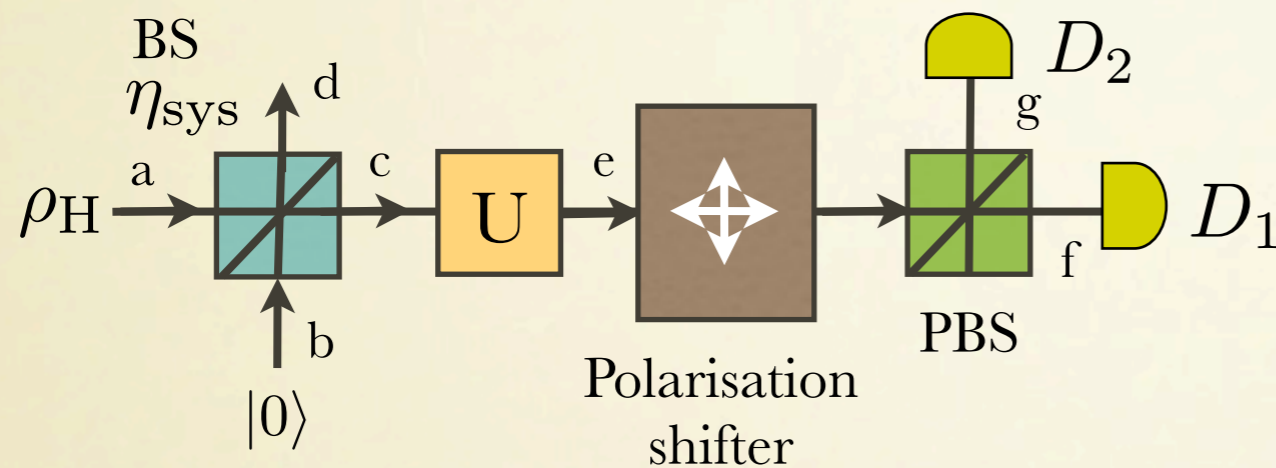


Misalignment in the channel:

$$\begin{pmatrix} c_H^\dagger \\ c_V^\dagger \end{pmatrix} = \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix} \begin{pmatrix} e_H^\dagger \\ e_V^\dagger \end{pmatrix}$$

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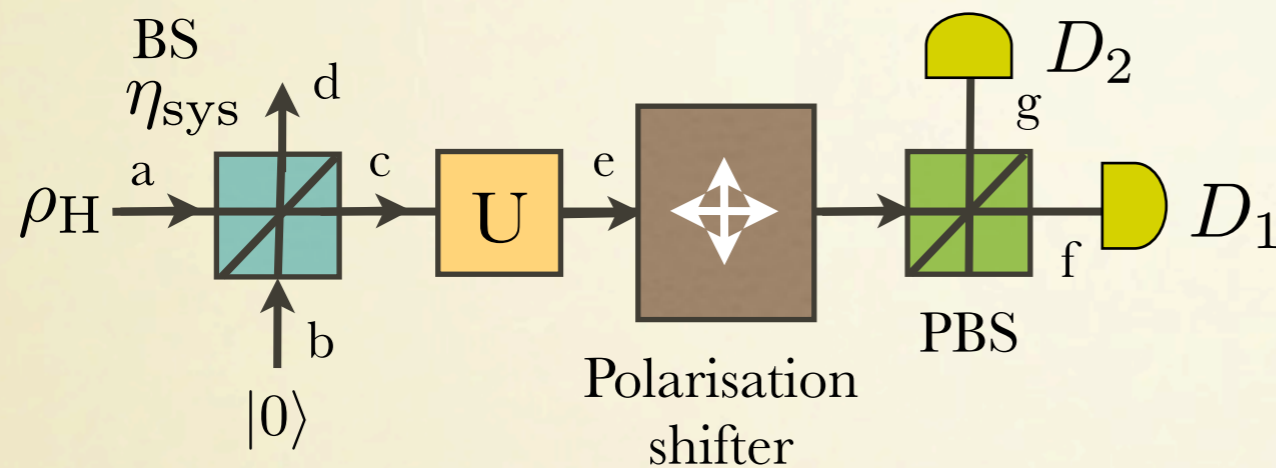
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The error rate can be written as:
$$E = \frac{1}{Q} e^{-\mu} \sum_{n=0}^{\infty} \frac{\mu^n}{n!} Y_n e_n$$

$Y_n e_n$: Probability that a n-photon signal produces a detected event associated with an error

CHARACTERISATION OF PRACTICAL DEVICES

Example: Error rate



Misalignment in the channel:

$$\begin{pmatrix} c_H^\dagger \\ c_V^\dagger \end{pmatrix} = \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix} \begin{pmatrix} e_H^\dagger \\ e_V^\dagger \end{pmatrix}$$

The error rate can be written as: $E = \frac{1}{Q} e^{-\mu} \sum_{n=0}^{\infty} \frac{\mu^n}{n!} Y_n e_n$

$Y_n e_n$: Probability that a n-photon signal produces a detected event associated with an error

$$Y_n e_n = \text{Tr} \left[\left(D_{1,\text{noclick}} \otimes D_{2,\text{click}} \otimes 1_d + \frac{1}{2} D_{1,\text{click}} \otimes D_{2,\text{click}} \otimes 1_d \right) |n\rangle_{dfg} \langle n| \right]$$



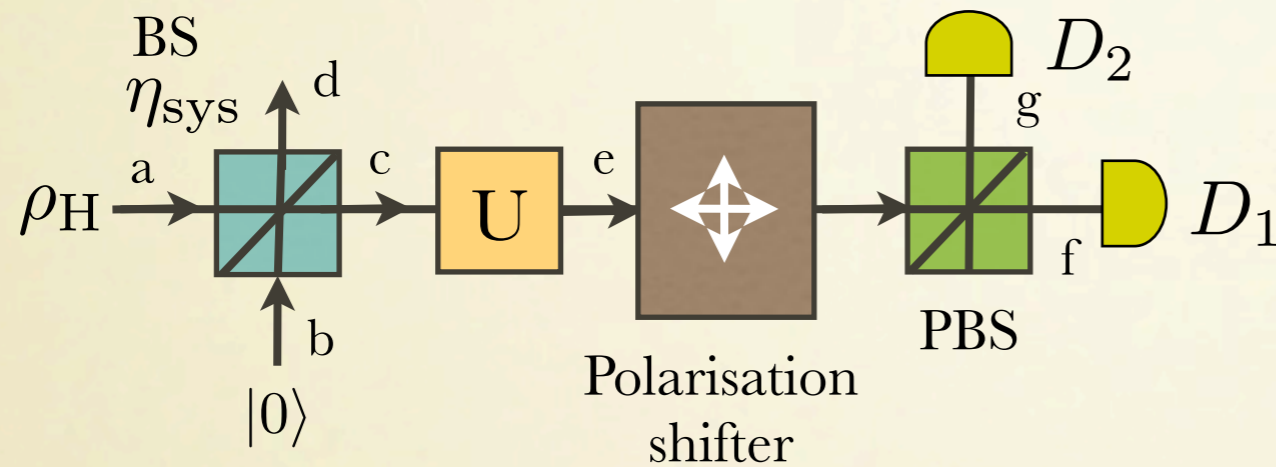
Double clicks are associated to random single clicks

CHARACTERISATION OF PRACTICAL DEVICES

Now we calculate: $|n\rangle_{dfg}$

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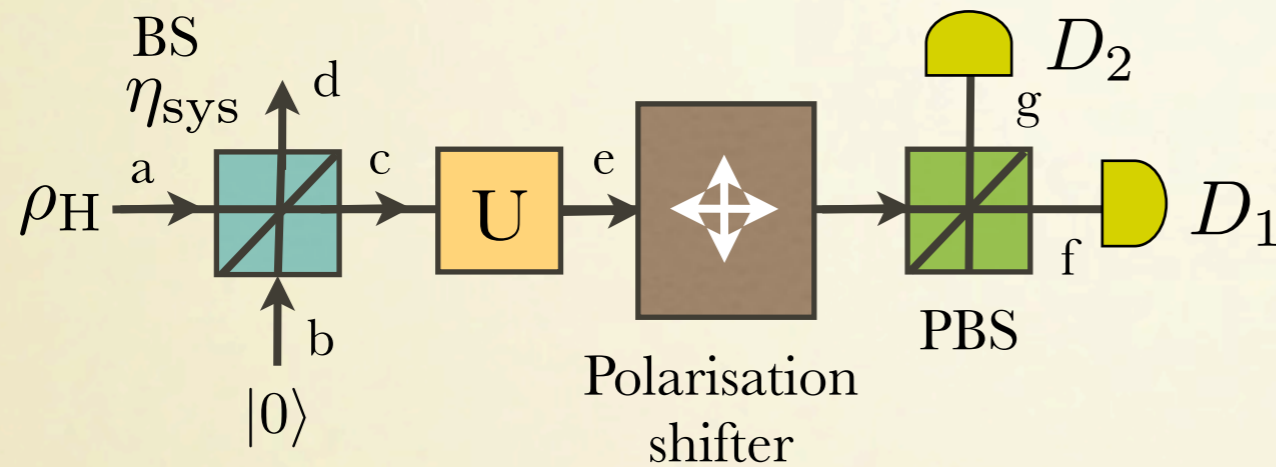


Input state: $\rho_H = |n\rangle\langle n|_H$
with

$$|n\rangle_H = \frac{1}{\sqrt{n!}} (a_H^\dagger)^n |0\rangle$$

CHARACTERISATION OF PRACTICAL DEVICES

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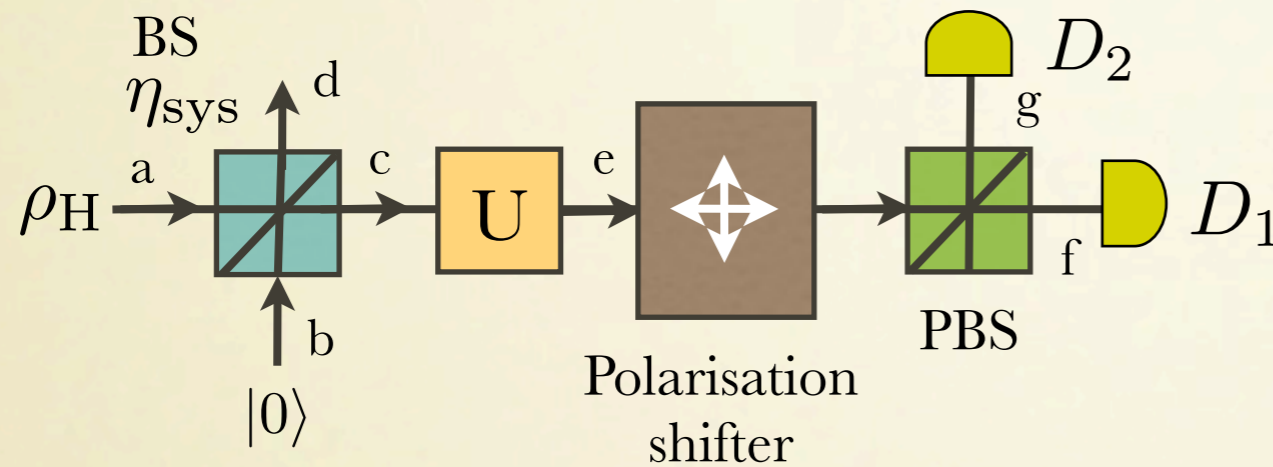
$$|n\rangle_H = \frac{1}{\sqrt{n!}} (a_H^\dagger)^n |0\rangle$$

$$a_H^\dagger \xrightarrow{\text{BS}} \sqrt{\eta_{\text{sys}}} c_H^\dagger + \sqrt{1 - \eta_{\text{sys}}} d_H^\dagger \xrightarrow{\text{U}} \sqrt{\eta_{\text{sys}}} (\cos \theta e_H^\dagger - \sin \theta e_V^\dagger) + \sqrt{1 - \eta_{\text{sys}}} d_H^\dagger$$

$$\xrightarrow{\text{PBS}} \sqrt{\eta_{\text{sys}}} (\cos \theta f_H^\dagger - \sin \theta g_V^\dagger) + \sqrt{1 - \eta_{\text{sys}}} d_H^\dagger$$

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$$|n\rangle_{dfg} = \sum_{k=0}^n \sum_{l=0}^{n-k} \sqrt{\frac{n!}{k!l!(n-k-l)!}} \sqrt{\eta_{\text{sys}}}^{n-k} \sqrt{1 - \eta_{\text{sys}}}^k (\cos \theta)^{n-k-l} (-\sin \theta)^l |k, n-k-l, l\rangle_{d_H, f_H, g_V}$$

CHARACTERISATION OF PRACTICAL DEVICES

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$$Y_n e_n = \text{Tr} \left[\left(D_{1,\text{noclick}} \otimes D_{2,\text{click}} \otimes 1_d + \frac{1}{2} D_{1,\text{click}} \otimes D_{2,\text{click}} \otimes 1_d \right) |n\rangle_{dfg} \langle n| \right]$$

$\longleftarrow \hspace{10em} \longrightarrow$
 D



$$D_{\text{noclick}} = (1 - d_{\text{dark}}) |0\rangle \langle 0|$$

$$D_{\text{click}} = 1 - D_{\text{noclick}}$$

CHARACTERISATION OF PRACTICAL DEVICES

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$$D = \frac{1}{2} \left[1_{dfg} + (1 - p_{\text{dark}}) (1_d \otimes |0\rangle\langle 0|_f \otimes 1_g - 1_d \otimes 1_f \otimes |0\rangle\langle 0|_g) - (1 - p_{\text{dark}})^2 (1_d \otimes |0\rangle\langle 0|_f \otimes |0\rangle\langle 0|_g) \right]$$

CHARACTERISATION OF PRACTICAL DEVICES

$$Y_n e_n = \text{Tr} \left[\left(\underbrace{D_{1,\text{noclick}} \otimes D_{2,\text{click}} \otimes 1_d + \frac{1}{2} D_{1,\text{click}} \otimes D_{2,\text{click}} \otimes 1_d}_D \right) |n\rangle_{dfg} \langle n| \right]$$

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We obtain:

$$Y_n e_n = \frac{1}{2} \left\{ 1 + (1 - p_{\text{dark}}) \frac{1}{2^n} [(2 - \eta_{\text{sys}} - \eta_{\text{sys}} \cos 2\theta)^n - (2 - \eta_{\text{sys}} + \eta_{\text{sys}} \cos 2\theta)^n] - (1 - p_{\text{dark}})^2 (1 - \eta_{\text{sys}})^n \right\}$$

CHARACTERISATION OF PRACTICAL DEVICES

$$E = \frac{1}{Q} e^{-\mu} \sum_{n=0}^{\infty} \frac{\mu^n}{n!} Y_n e_n$$
$$= \frac{1}{2Q} \left[1 + (1 - p_{\text{dark}}) \left(e^{-\mu\eta_{\text{sys}} \cos^2 \theta} - e^{-\mu\eta_{\text{sys}} \sin^2 \theta} \right) - (1 - p_{\text{dark}})^2 e^{-\mu\eta_{\text{sys}}} \right]$$

The error rate is directly observed in the experiment.

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Example: BB84 with phase-randomised WCPs

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Example: BB84 with phase-randomised WCPs

$$R \geq q \{ p_1 Y_1 [1 - h(e_1)] - Q h(E) \}$$

q	is the basis-sift factor (known)
$p_1 = \mu e^{-\mu}$	is the probability that Alice emits a single-photon state (known)
Y_1	is the yield of the single-photon states (unknown)
e_1	is the phase error of the single photon states (unknown)
Q	is the overall gain of the signal states (observed)
E	is the overall error rate of the signal states (observed)

D. Gottesman, H.-K. Lo, N. Lütkenhaus and J. Preskill, Quantum Inf. Comput. 4, 325 (2004).

CHARACTERISATION OF PRACTICAL DEVICES

We assume that Q, E , is the same for both basis. Parameter estimation (due to the PNS attack we need to consider the worst-case scenario):

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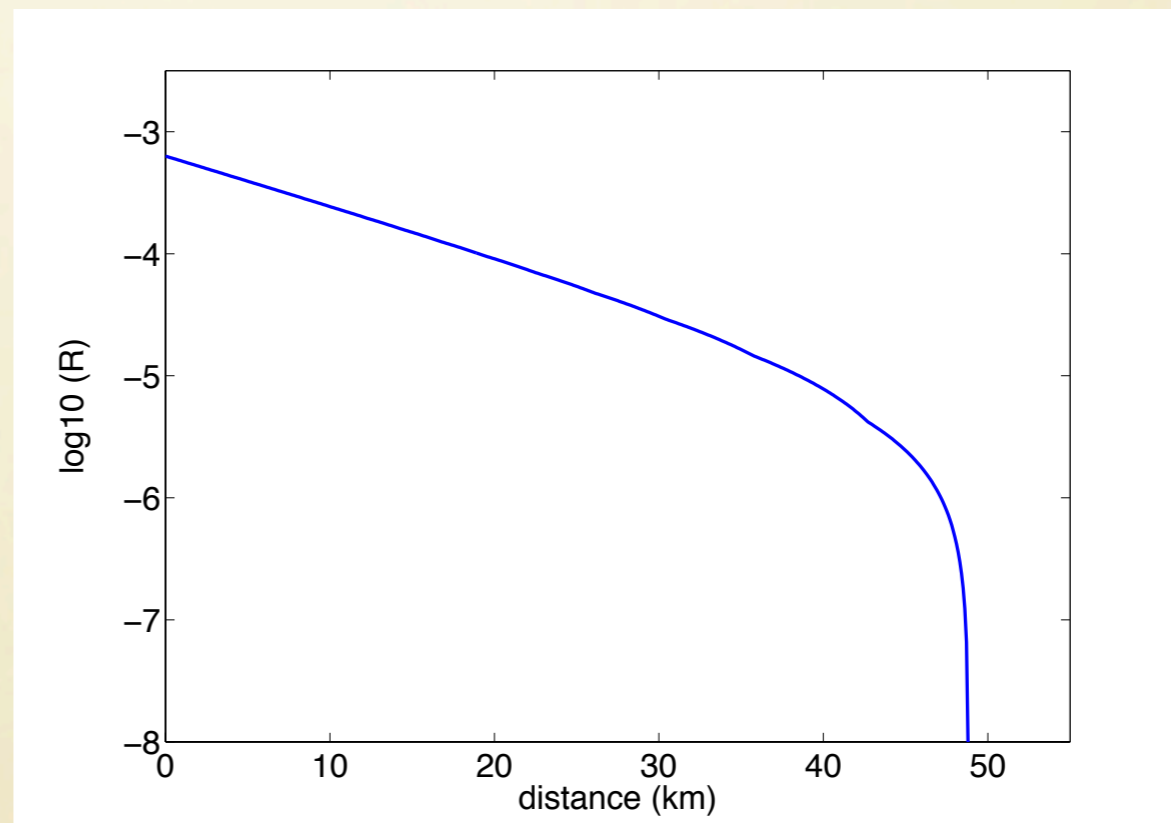
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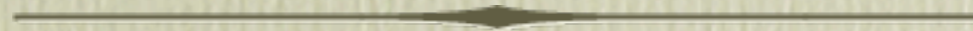
$$q \approx 1$$

$$\mu \text{ optimised}$$



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QKD with decoy states (asymptotic case)

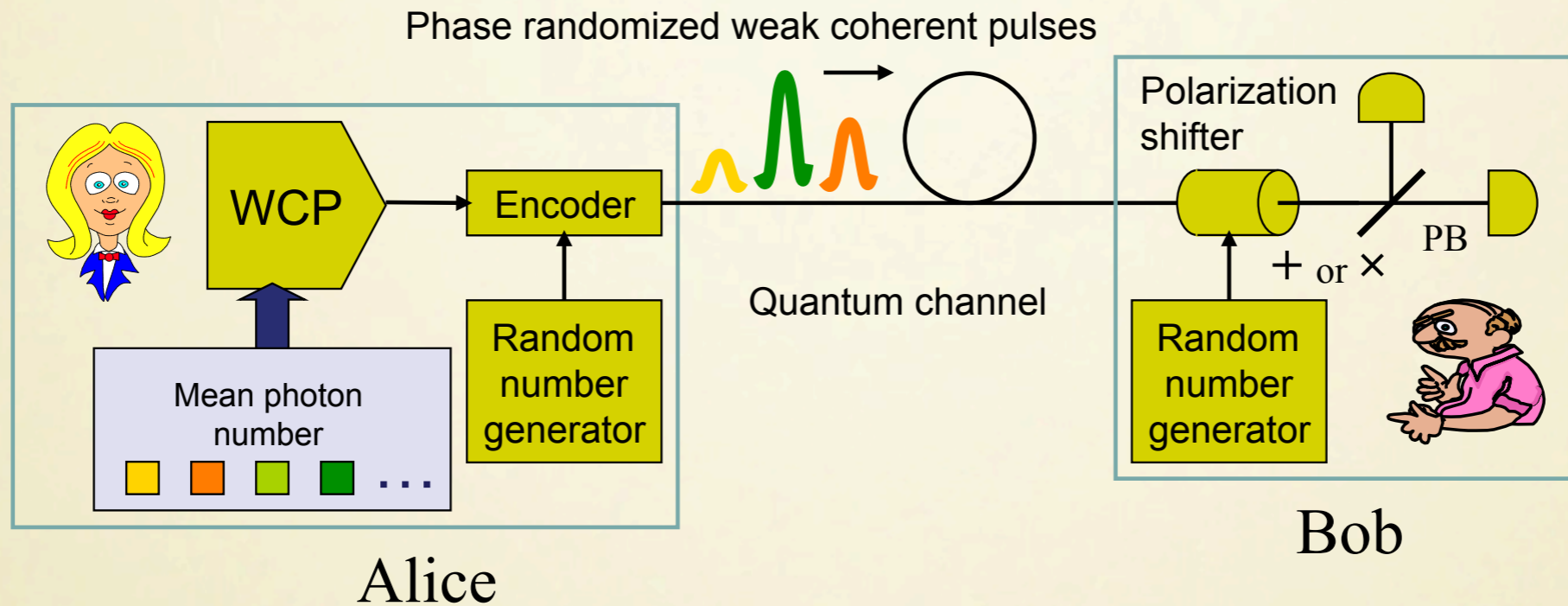


QKD WITH DECOY STATES

Motivation: Better estimation of Y_1, e_1 .

QKD WITH DECOY STATES

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Alice prepares phase-randomised weak coherent pulses whose mean photon number is chosen for each signal from a finite set of possible values.

$$\rho_l = e^{-\mu_l} \sum_{n=0}^{\infty} \frac{\mu_l^n}{n!} |n\rangle \langle n| \quad \text{with } l \in \{s, d_1, d_2, \dots, d_N\}$$

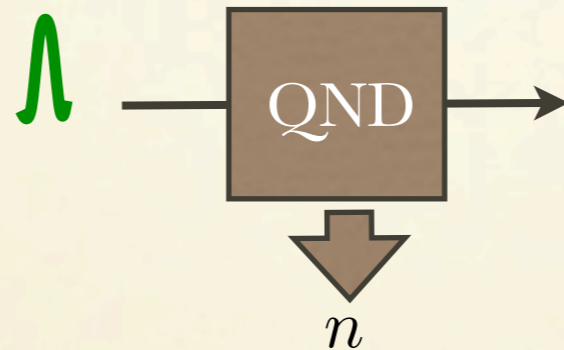
W.-Y. Hwang, PRL 91, 057901 (2003); H.-K. Lo, X. Ma and K. Chen, PRL 94, 230504 (2005); X.-B. Wang, PRL 94, 230503 (2005).

QKD WITH DECOY STATES

Intuition:

In principle Eve can guess the intensity setting l selected by Alice:

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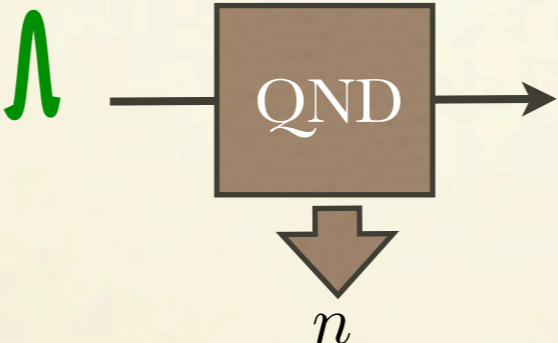
$$\begin{aligned} p(l|n) &= p(n|l) \frac{p(l)}{p(n)} \\ &= e^{\mu_l} \frac{\mu_l^n}{n!} \frac{p(l)}{\sum_l p(l) e^{\mu_l} \mu_l^n / n!} \end{aligned}$$

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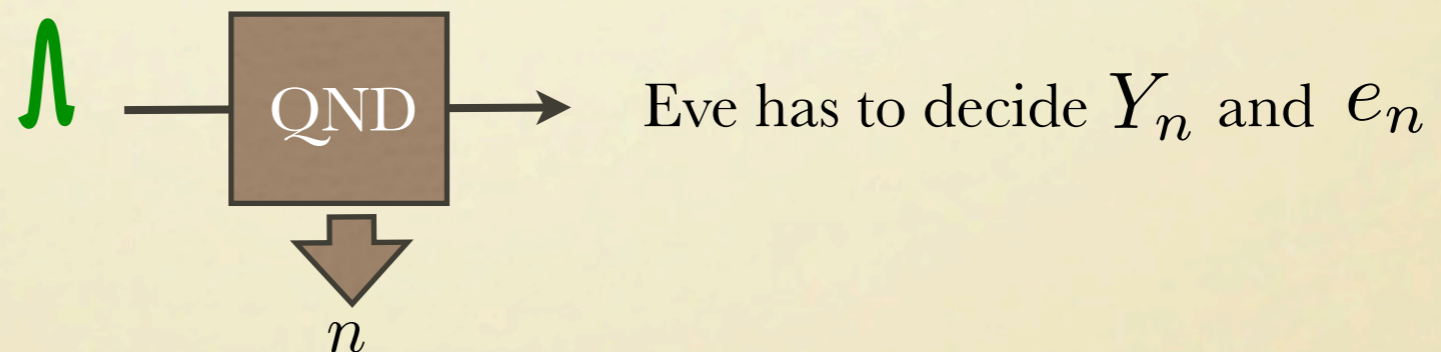
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$$= e^{\mu_l} \frac{\mu_l^n}{n!} \frac{p(l)}{\sum_l p(l) e^{\mu_l} \mu_l^n / n!}$$

Key idea: The yields Y_n and the error rates e_n are equal for the different intensity settings



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$$Q_{d_1} = e^{-\mu_{d_1}} \sum_{n=0}^{\infty} \frac{\mu_{d_1}^n}{n!} Y_n$$

$$E_{d_1} Q_{d_1} = e^{-\mu_{d_1}} \sum_{n=0}^{\infty} \frac{\mu_{d_1}^n}{n!} Y_n e_n$$

⋮

⋮

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where \mathbf{x} is a vector of unknown variables, c and b are vectors whose coefficients are known, and A is a known matrix.

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$$Q_l \geq e^{-\mu_l} \sum_{n=0}^{M_{\text{cut}}} \frac{\mu_l^n}{n!} Y_n$$

$$Q_l \leq e^{-\mu_l} \sum_{n=0}^{M_{\text{cut}}} \frac{\mu_l^n}{n!} Y_n + e^{-\mu_l} \sum_{n=M_{\text{cut}}+1}^{\infty} \frac{\mu_l^n}{n!} = e^{-\mu_l} \sum_{n=0}^{M_{\text{cut}}} \frac{\mu_l^n}{n!} Y_n + \left(1 - e^{-\mu_l} \sum_{n=0}^{M_{\text{cut}}} \frac{\mu_l^n}{n!} \right)$$

QKD WITH DECOY STATES

$$\min Y_1$$

$$\text{s.t. } Q_l \geq e^{-\mu_l} \sum_{n=0}^{M_{\text{cut}}} \frac{\mu_l^n}{n!} Y_n \quad \forall l$$

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$$1 \geq Y_n \geq 0$$



Lower bound
for Y_1

This is done for both BB84 basis.

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Similarly, if we define $\gamma_n = Y_n e_n$

$$\max \gamma_1$$

$$\text{s.t. } E_l Q_l \geq e^{-\mu_l} \sum_{n=0}^{M_{\text{cut}}} \frac{\mu_l^n}{n!} \gamma_n \quad \forall l$$

$$E_l Q_l \leq e^{-\mu_l} \sum_{n=0}^{M_{\text{cut}}} \frac{\mu_l^n}{n!} \gamma_n + \left(1 - e^{-\mu_l} \sum_{n=0}^{M_{\text{cut}}} \frac{\mu_l^n}{n!} \right) \quad \forall l$$

$$1 \geq \gamma_n \geq 0$$



Upper bound
for e_1 :

$$e_1 \leq \frac{\gamma_1}{Y_1}$$

QKD WITH DECOY STATES

$$R \geq q \{ p_{1|s} Y_1 [1 - h(e_1)] - Q_s h(E_s) \}$$

- $p_{1,s} = \mu_s e^{-\mu_s}$ is the conditional probability that Alice emits a single-photon state when she uses the signal intensity setting (**known**)
- Q_s is the overall gain of the signal states (**observed**)
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If we use the channel model described before:

Example:

$$p_{\text{dark}} = 10^{-6}$$

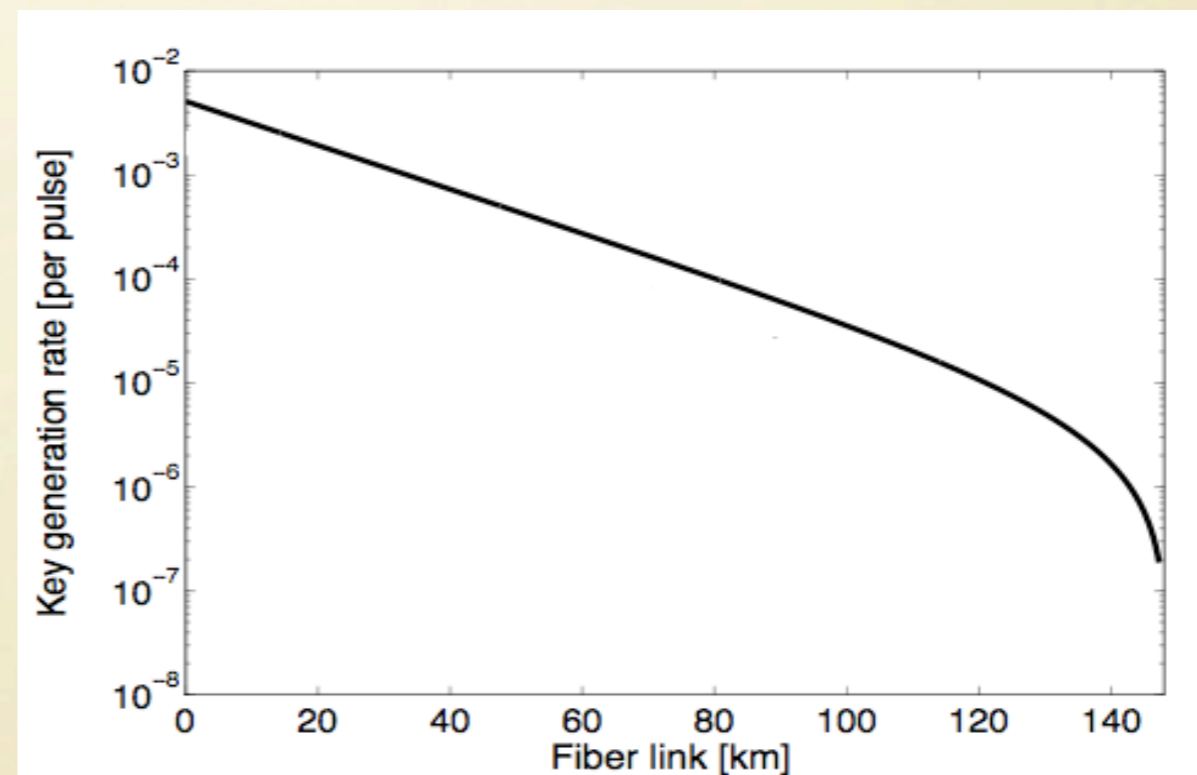
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Parameter estimation (finite case)

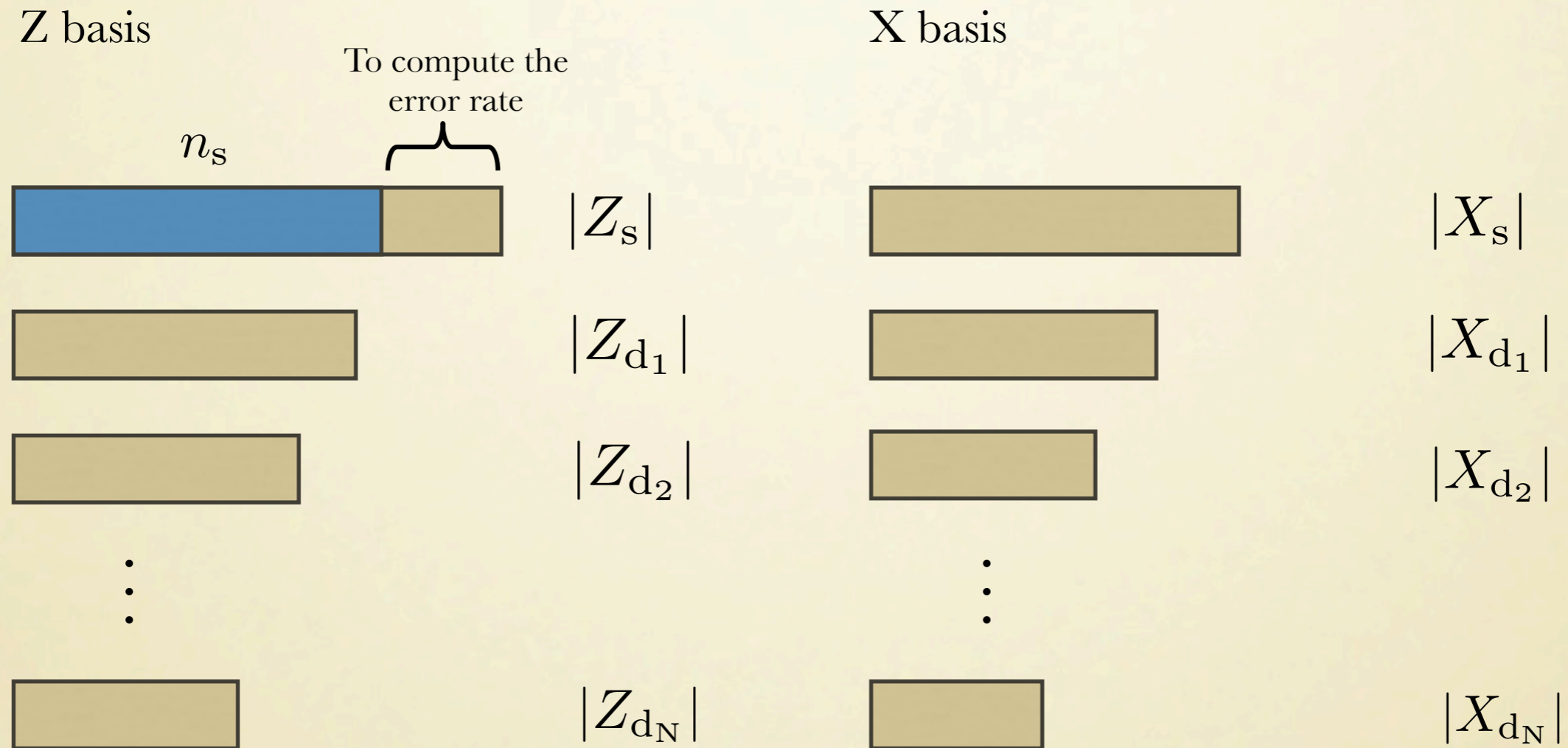


PARAMETER ESTIMATION (FINITE CASE)

In any experiment Alice only sends a finite number of signals. When the sifting conditions are met we have that

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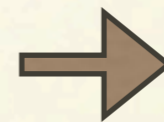
We need to compute a lower bound for the number of single photons and an upper bound for their phase error rate in the set

PARAMETER ESTIMATION (FINITE CASE)

Actual protocol (let us focus, for instance, in the Z basis):



Alice chooses an intensity setting l with probability $p(l|Z)$



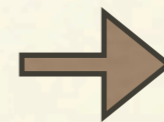
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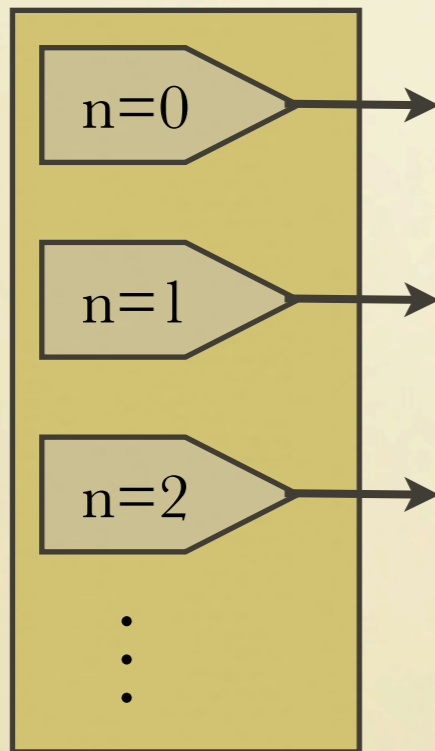


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Equivalent protocol:



For each signal, Alice first chooses a photon number n with probability

$$p(n|Z) = \sum_l p(l|Z)p(n|l, Z)$$

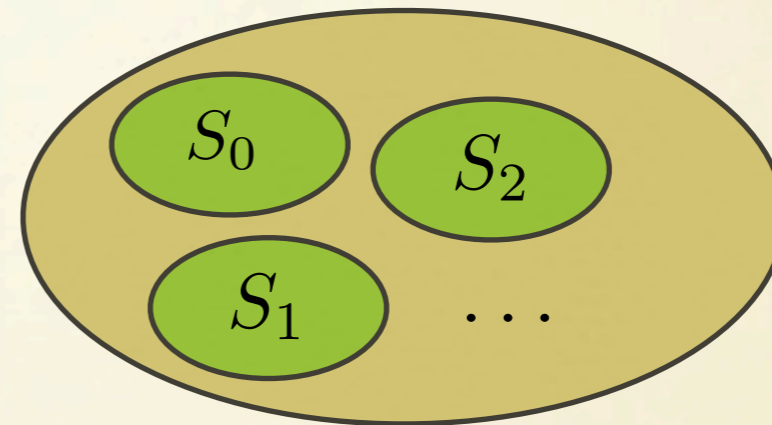
After Bob declares the detected events, Alice decides the intensity setting l with probability

$$p(l|n, Z) = p(n|l, Z) \frac{p(l|Z)}{p(n|Z)}$$

PARAMETER ESTIMATION (FINITE CASE)

Let S_n denote the number of signals sent by Alice with n photons, when both Alice and Bob select the basis Z , and Bob obtains a click in his measurement apparatus.

$$\sum_l |Z_l| = \sum_n S_n$$

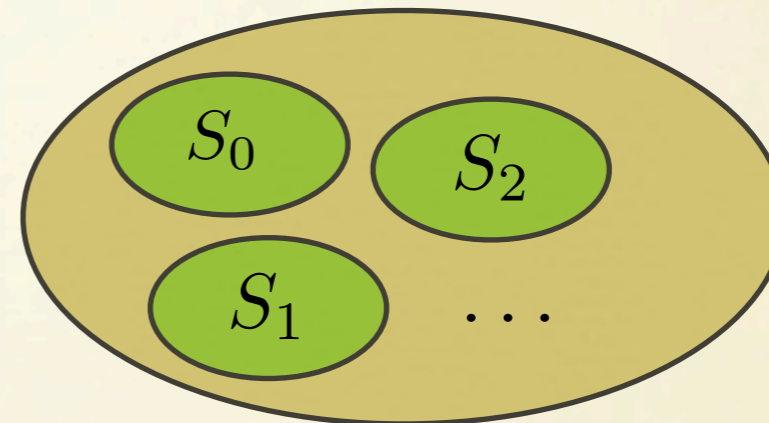


Set of detected events

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Set of detected events

Using the equivalent protocol we expect to be able to write:

$$|Z_l| = \sum_n p(l|n, Z) S_n + \delta_l$$

⏟
⏟
⏟
⏟

observed *known* *unknown* *can be bounded*

We will be able to obtain the parameters S_n , in particular S_1

PARAMETER ESTIMATION (FINITE CASE)

How to bound the fluctuation term $\delta_l \rightarrow$ Example: Chernoff bound

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How to bound the fluctuation term $\delta_l \rightarrow$ Example: Chernoff bound

Claim 1. Let X_1, X_2, \dots, X_n , be a set of independent Bernoulli trials that satisfy $\Pr(X_i = 1) = p_i$. And, let $X = \sum_{i=1}^n X_i$ and $\mu = E[X] = \sum_{i=1}^n p_i$, where $E[\cdot]$ is the mean value. Then, we have that

$$X = \mu + \delta, \tag{B1}$$

except with error probability $\gamma = \varepsilon + \hat{\varepsilon}$, where the parameter $\delta \in [-\Delta, \hat{\Delta}]$, with $\Delta = g(X, \varepsilon^{2(4+\sqrt{7})^2/9})$ and $\hat{\Delta} = g(X, \hat{\varepsilon}^3)$, and the function $g(x, y) = \sqrt{x \ln(y^{-1})}$, given that $\max\{\hat{\varepsilon}^{-1/X}, \varepsilon^{-1/X}\} \leq \exp(1/3)$.

PARAMETER ESTIMATION (FINITE CASE)

How to bound the fluctuation term $\delta_l \rightarrow$ Example: Chernoff bound

Claim 1. Let X_1, X_2, \dots, X_n , be a set of independent Bernoulli trials that satisfy $\Pr(X_i = 1) = p_i$. And, let $X = \sum_{i=1}^n X_i$ and $\mu = E[X] = \sum_{i=1}^n p_i$, where $E[\cdot]$ is the mean value. Then, we have that

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This implies that

$$|Z_l| = \sum_n p(l|n, Z) S_n + \delta_l$$

except with error probability $\gamma_l = \varepsilon_l + \hat{\varepsilon}_l$, where $\delta_l \in [-\Delta_l, \hat{\Delta}_l]$, with

$$\Delta_l = g(|Z_l|, \varepsilon_l^{2(4+\sqrt{7})^2/9})$$

$$\hat{\Delta}_l = g(|Z_l|, \hat{\varepsilon}_l^3)$$



Importantly, the fluctuation term is bounded by observed quantities and the tolerated failure probability

PARAMETER ESTIMATION (FINITE CASE)

We have **more conditions**: $N_n \geq S_n \geq 0$

N_n : Number of signals sent by Alice with n photons, when she and Bob select the Z basis.

PARAMETER ESTIMATION (FINITE CASE)

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Using Chernoff inequality, we have that

$$p(N_n \geq N[p(n|Z) + \xi_n]) \leq e^{-N\xi_n^2/[2(p(n|Z) + \xi_n)]}$$

$$p(N_n \leq N[p(n|Z) - \xi_n]) \leq e^{-N\xi_n^2/[2p(n|Z)]}$$

where $N = \sum_n N_n$ is the number of signals sent by Alice and measured by Bob in the Z basis

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Equivalently, we can say that $N_n = N[p(n|Z) + \delta_n]$

except with error probability $\gamma_n = \epsilon_n + \hat{\epsilon}_n$, where $\delta_n \in [-\Delta_n, \hat{\Delta}_n]$, with

$$\Delta_n = \min \{g[p(n|Z)/N, \epsilon_n^2], p(n|Z)\}$$

$$\hat{\Delta}_n = \min \{f[N, p(n|Z), \hat{\epsilon}_n], 1 - p(n|Z)\}$$

← We also use $N \geq N_n \geq 0$

where $g(x, y) = \sqrt{x \ln(y^{-1})}$ and $f(x, y, z) = \ln(z^{-1})[1 + \sqrt{1 + 2xy/\ln(z^{-1})}]/x$

PARAMETER ESTIMATION (FINITE CASE)

Based on the foregoing:

$$\min S_1$$

$$\text{s.t. } |Z_l| = \sum_{n=0}^{\infty} p(l|n, \mathbf{Z}) S_n + \delta_l, \quad \forall l$$

$$\hat{\Delta}_l \geq \delta_l \geq -\Delta_l, \quad \forall l$$

$$\sum_l \delta_l = 0, \quad \forall l \quad (\text{from the condition } \sum_l |Z_l| = \sum_n S_n)$$

$$N[p(n|\mathbf{Z}) + \delta_n] \geq S_n \geq 0, \quad \forall n$$

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This linear optimisation problem can be solved analytically or numerically using linear programming

PARAMETER ESTIMATION (FINITE CASE)

Example: Solution using **linear programming**. We reduce the number of unknown parameters to a finite set:

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Example: Solution using **linear programming**. We reduce the number of unknown parameters to a finite set:

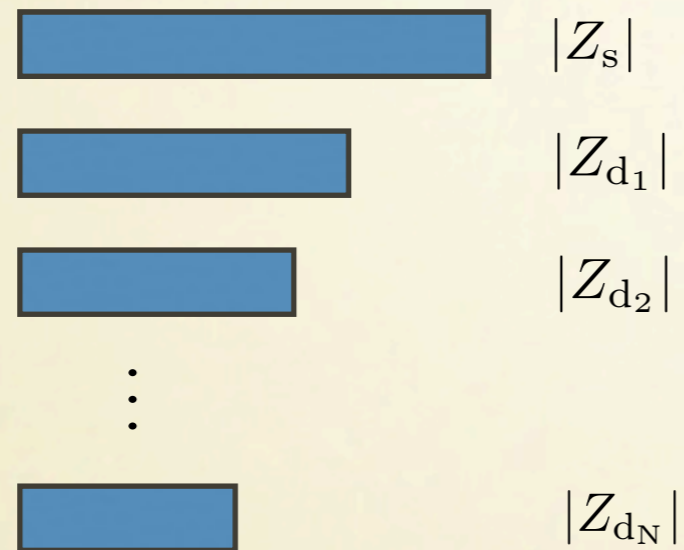
$$\begin{aligned}
 & \min S_1 \\
 & \text{s.t. } |Z_l| \geq \sum_{n \in \mathcal{S}_{\text{cut}}} p(l|n, Z) S_n + \delta_l, \quad \forall l \\
 & |Z_l| \leq \sum_{n \in \mathcal{S}_{\text{cut}}} p(l|n, Z) S_n + \delta_l + \max_{j \notin \mathcal{S}_{\text{cut}}} p(l|j, Z) N \left[1 - \sum_{n \in \mathcal{S}_{\text{cut}}} (p(n|Z) + \delta_n) \right], \quad \forall l \\
 & \hat{\Delta}_l \geq \delta_l \geq -\Delta_l, \quad \forall l \\
 & \sum_l \delta_l = 0, \quad \forall l \\
 & N[p(n|Z) + \delta_n] \geq S_n \geq 0, \quad \forall n \in \mathcal{S}_{\text{cut}} \\
 & \hat{\Delta}_n \geq \delta_n \geq -\Delta_n, \quad \forall n \in \mathcal{S}_{\text{cut}}
 \end{aligned}$$

except with error probability ϵ_1 given by $\epsilon_1 \leq \sum_l \gamma_l + \sum_{n \in \mathcal{S}_{\text{cut}}} \gamma_n$

Here: $\mathcal{S}_{\text{cut}} = \{n : 0 \leq n \leq M_{\text{cut}}\}$

PARAMETER ESTIMATION (FINITE CASE)

S_1 is a lower bound for the number of single photon in the Z basis:



PARAMETER ESTIMATION (FINITE CASE)

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PARAMETER ESTIMATION (FINITE CASE)

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Using again Chernoff bound: $n_1 \geq p(s|1, \mathbf{Z}) \frac{n_s}{|Z_s|} S_1 - \Delta_1$

except with error probability ϵ'_1 , where:

$$\Delta_1 = g \left(p(s|1, \mathbf{Z}) \frac{n_s}{|Z_s|} S_1, \epsilon_1'^2 \right)$$

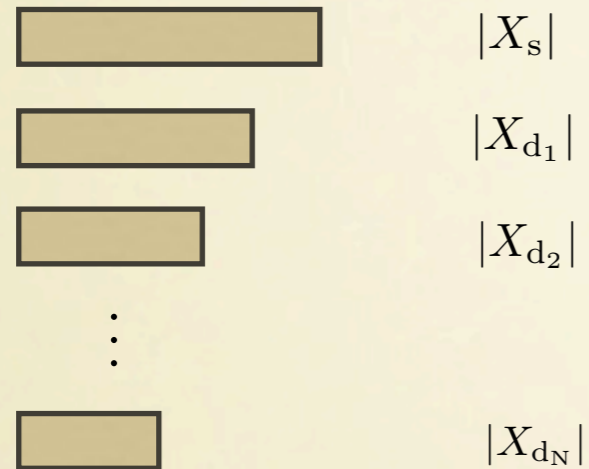
Total error probability in the estimation of n_1 : $\epsilon_1 \leq \epsilon'_1 + \sum_l \gamma_l + \sum_{n \in \mathcal{S}_{\text{cut}}} \gamma_n$

PARAMETER ESTIMATION (FINITE CASE)

Let us know calculate the phase error of the single photons:

PARAMETER ESTIMATION (FINITE CASE)

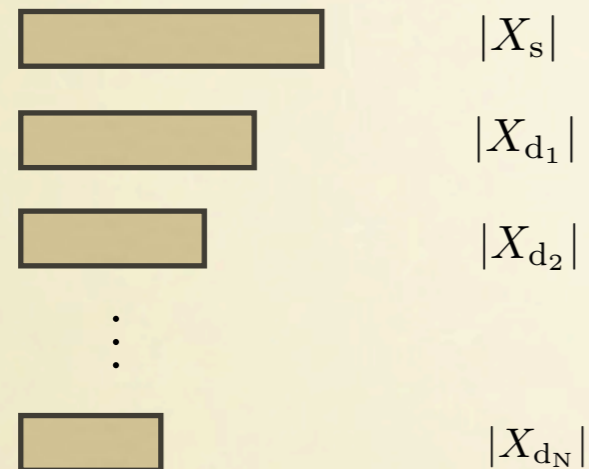
Let us know calculate the phase error of the single photons:



Using the same techniques as before we can obtain a lower bound for S_1 (in the \mathbf{X} basis) and an upper bound for the number of errors \bar{e}_1 associated to single-photon events in the \mathbf{X} basis

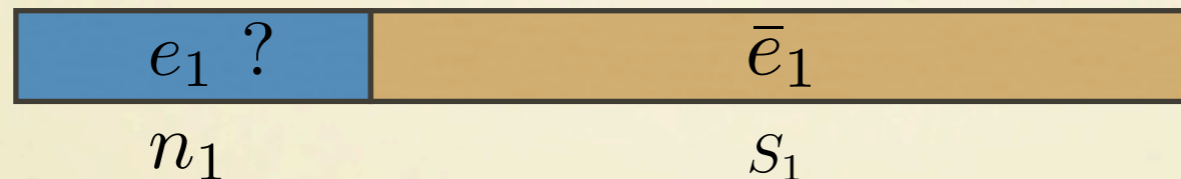
PARAMETER ESTIMATION (FINITE CASE)

Let us now calculate the phase error of the single photons:



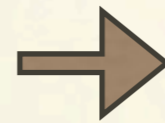
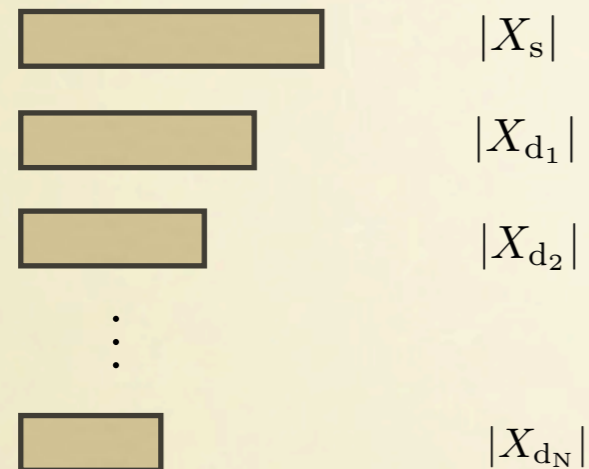
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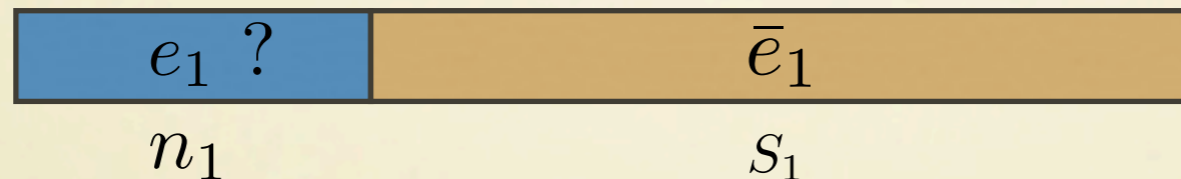
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Now we can use a result from random sampling without replacement:



$$e_1 \leq \min \left\{ \left[n_1 \left(\frac{\bar{e}_1}{S_1} \right) + (n_1 + S_1) \Omega(n_1, S_1, \epsilon_e) \right], n_1/2 \right\} \text{ with } \Omega(x, y, z) = \sqrt{(x+1) \ln(z^{-1}) / (2y(x+y))}$$

except with error probability $\epsilon_{e_1} \leq \epsilon_e + \sum_l (\gamma_l + \gamma_{l,e}) + \sum_{n \in \mathcal{S}_{\text{cut}}} \gamma_n$

R. J. Serfling, Ann. Statist. 2 (1), 39-48 (1974).

Side-channels



SIDE-CHANNELS

Are experimental implementations of QKD really secure?

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Are experimental implementations of QKD really secure?

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News

Quantum crack in cryptographic armour

A commercial quantum cryptography system has been hacked for the first time.

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Jun 17, 2011 | 5 comments

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Jul 26th 2010

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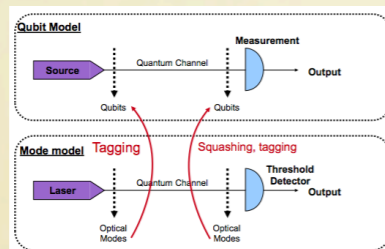
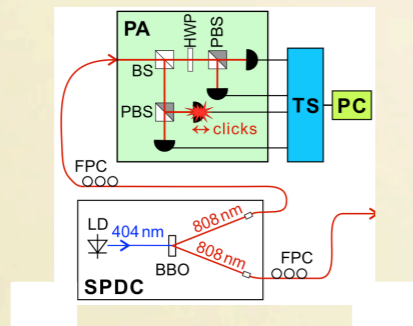
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Quantum crack in cryptographic armour

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The security proof of a QKD system typically includes several steps

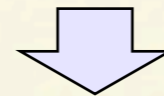


$$\frac{1}{2} \|\rho_{AE} - U_A \otimes \rho_E\| \leq \epsilon$$

$$\rho_{AE} = \sum_s |s\rangle\langle s| \otimes \rho_E^s$$

$$U_A = \frac{1}{|S|} \sum_s |s\rangle\langle s|$$

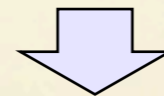
Actual physical devices



Quantum optical model
e.g. mode based

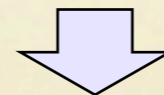
Modelling

e.g. realistic laser sources
beamsplitters model
threshold detectors model



Security model
e.g. qubit based

Reduction to essentials
e.g. tagging, squashing



Security proof

Entanglement distillation
Information theoretic

From a mathematical model for employed devices we can provide a scientific (mathematical and physical) universally composable security proof for QKD: perfect key except with probability ϵ

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Modelling of real devices: What can go wrong?

SIDE-CHANNELS

Modelling of real devices: What can go wrong?

State preparation:

- Does the source emit coherent states?
- Are the states truly phase-randomised?
- Are we preparing perfect BB84 states?
- Are the states single-mode?
- Consider intensity fluctuations in the source...

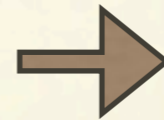
⋮

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-



If we know the imperfections we can include them in the security proof

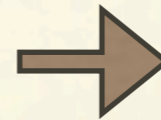


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Measurement device:

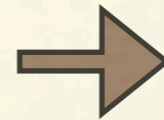
- Problem with efficiency mismatch
- Take into account the dead-time of the detectors
- Guarantee that the BS (passive receiver) cannot be controlled by Eve (e.g. wavelength dependence)
- ⋮
- **Do the detectors behave as we expect?**

SIDE-CHANNELS

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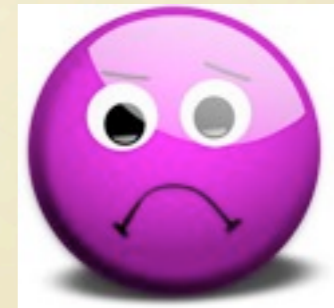


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- Take into account the dead-time of the detectors
- Guarantee that the BS (passive receiver) cannot be controlled by Eve (e.g. wavelength dependence)
- ⋮
- **Do the detectors behave as we expect?**




The **weakest link** in a QKD system is the measurement device

SIDE-CHANNELS

Quantum hacking: Blinding attack



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NATURE PHOTONICS | LETTER

Hacking commercial quantum cryptography systems by tailored bright illumination

Lars Lydersen, Carlos Wiechers, Christoffer Wittmann, Dominique Elser, Johannes Skaar & Vadim Makarov

[Affiliations](#) | [Contributions](#) | [Corresponding author](#)







Nature Photonics 4, 686–689 (2010) | doi:10.1038/nphoton.2010.214
Received 02 April 2010 | Accepted 11 July 2010 | Published online 29 August 2010

Abstract

[Abstract](#) • [Author information](#) • [Supplementary information](#)

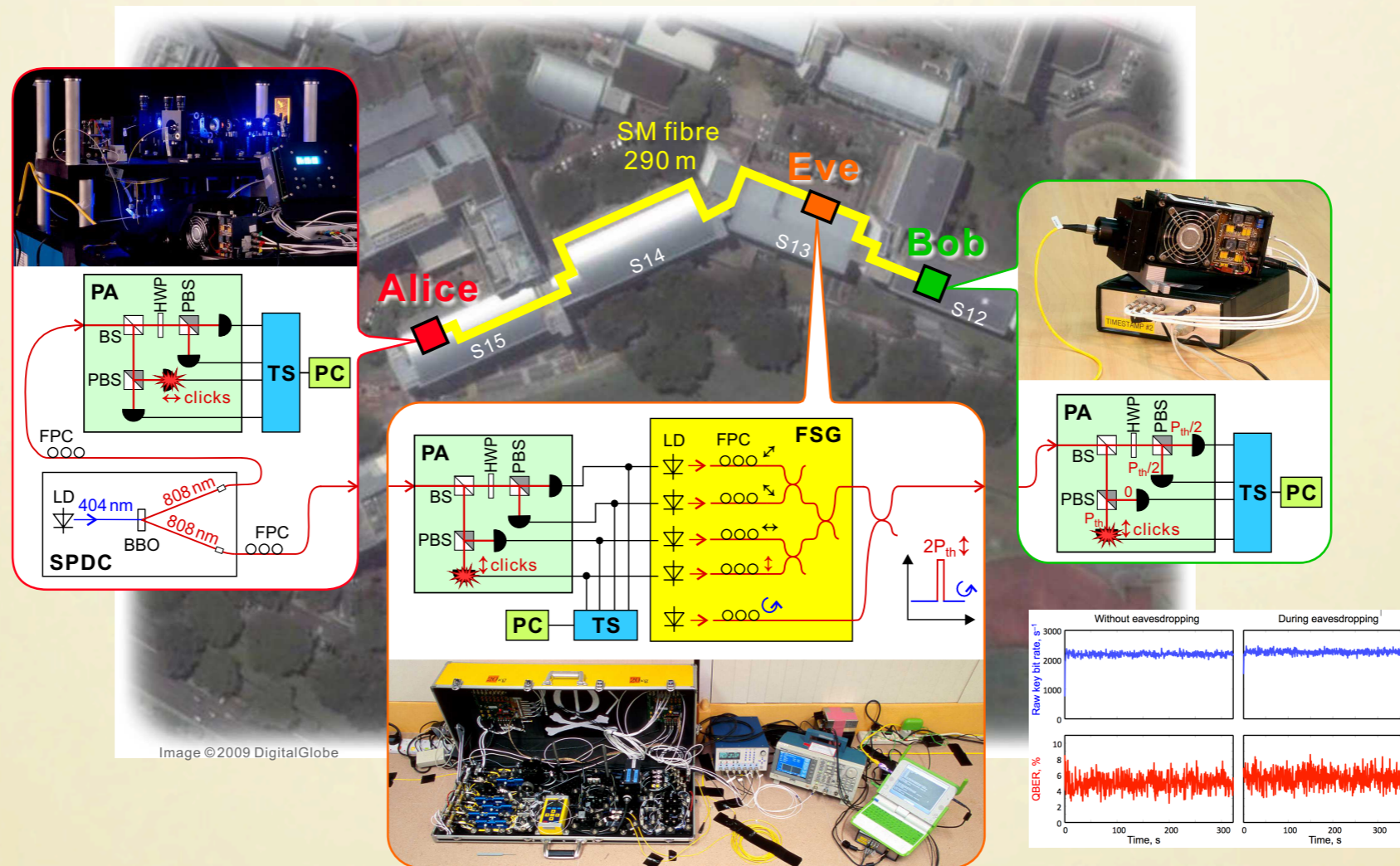
The peculiar properties of quantum mechanics allow two remote parties to communicate a private, secret key, which is protected from eavesdropping by the laws of physics^{1, 2, 3, 4}. So-called quantum key distribution (QKD) implementations always rely on detectors to measure the relevant quantum property of single photons⁵. Here we demonstrate experimentally that the detectors in two commercially available QKD systems can be fully remote-controlled using specially tailored bright illumination. This makes it possible to tracelessly acquire the full secret key; we propose an eavesdropping apparatus built from off-the-shelf components. The loophole is likely to be present in most QKD systems using avalanche photodiodes to detect single photons. We believe that our findings are crucial for strengthening the security of practical QKD, by identifying and patching technological deficiencies.

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Eavesdropping 100% of the key on installed QKD line.



I. Gerhardt et al., Nature Comm. 2, 349 (2011).

See also:

Y. Zhao et al., Phys. Rev. A 78, 042333 (2008).

N. Jain et al., Phys. Rev. Lett. 107, 110501 (2011).

H. Weier et al., New J. Phys. 13, 073024 (2011)

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Bridging the gap between theory and practice...

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Bridging the gap between theory and practice...

Option 1: "Patches"

- *Abandon the provable security model of QKD*
- *Can often be defeated by hacking advances*

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Bridging the gap between theory and practice...

Option 1: "Patches"

- *Abandon the provable security model of QKD*
- *Can often be defeated by hacking advances*

Option 2: Integrate imperfections into the security proof

- *Typically, it may need deep modification of the protocol, hardware and security proof*
- ***Device-independent quantum key distribution*** (avoids the hard-verifiable requirement of completely characterizing real devices)

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Device independent QKD (diQKD)/Self-testing QKD

D. Mayers and A. C.-C. Yao, in Proc. 39th Annual Symposium on Foundations of Computer Science (FOCS98), p. 503 (1998); A. Acín et al., Phys. Rev. Lett. 98, 230501 (2007); A. Acín, N. Gisin and Ll. Masanes, Phys. Rev. Lett. 97, 120405 (2006).

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Device independent QKD (diQKD)/Self-testing QKD



We still need some assumptions: validity of QM, true RNG, Alice and Bob shielded from Eve, no memory, ... Removes the problem of full characterising real devices!

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BASIC idea: The existence of entanglement => possibility of secure key generation

Bell inequalities test => Entanglement verification

Alice and Bob can perform Bell inequality test with untrusted devices

If $p(a,b|x,y)$ violates some Bell inequality, then $p(a,b|x,y)$ contains secrecy irrespectively of the implementation!

Advantage: diQKD eliminates ALL potential side-channels

D. Mayers and A. C.-C. Yao, in Proc. 39th Annual Symposium on Foundations of Computer Science (FOCS98), p. 503 (1998); A. Acín et al., Phys. Rev. Lett. 98, 230501 (2007); A. Acín, N. Gisin and Ll. Masanes, Phys. Rev. Lett. 97, 120405 (2006).

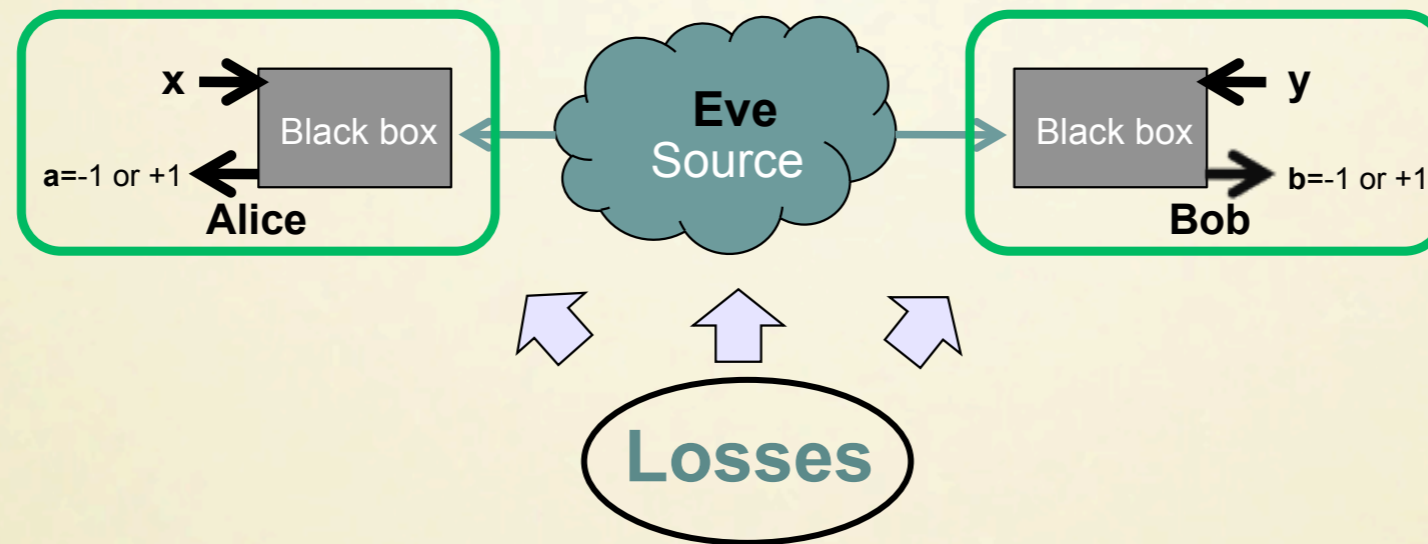
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Now... let's go to the lab

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Now... let's go to the lab

We need to violate a Bell inequality loophole-free \rightarrow Very hard!

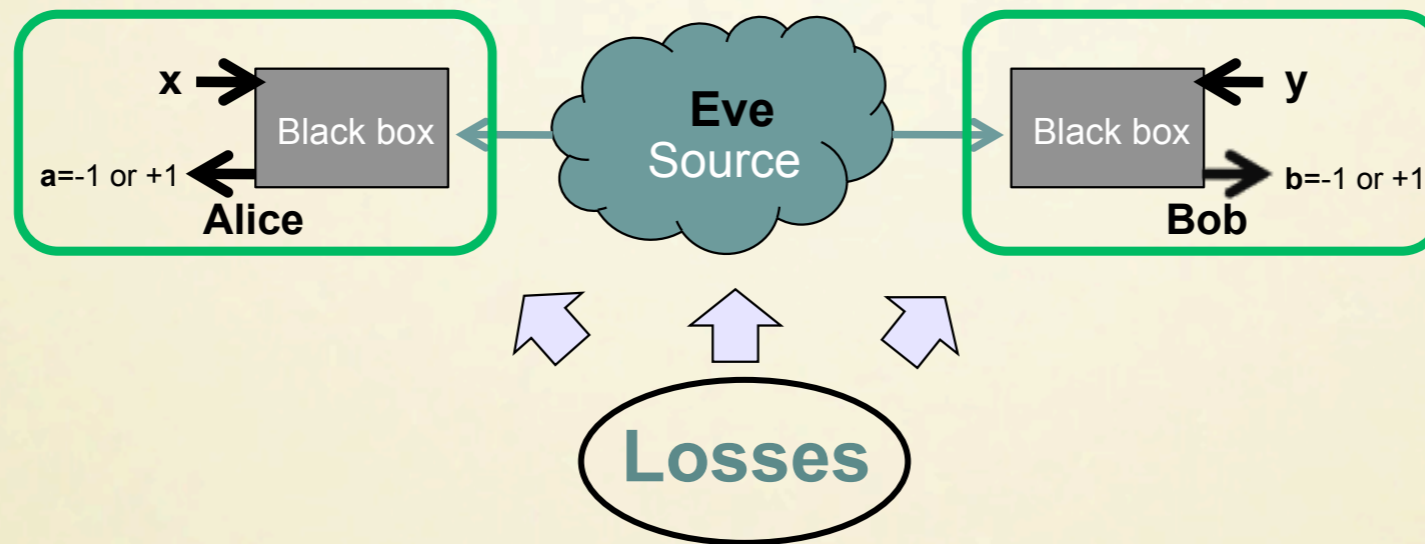


Patch: random/deterministic assignment for lost signals \rightarrow increase error rate \rightarrow loss of violation

SIDE-CHANNELS

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**Detection
loophole**

Required detection efficiency $> 82.8\%$

But the transmission efficiency of 5 km of telecom fiber is roughly 80%; typical detection efficiencies are 10-15%

SIDE-CHANNELS

Fair-sampling device

N. Gisin, S. Pironio and N. Sangouard, Phys. Rev. Lett. 105, 070501 (2010); N. Sangouard et al., Phys. Rev. Lett. 106, 120403 (2011); M. Curty and T. Moroder, Phys. Rev. A 84, 010304(R) (2011).

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Fair-sampling device

In Bell tests \rightarrow assume that the set of detected photon pairs is a fair set (fair-sampling assumption). It is reasonable to assume that Nature is not malicious.

In diQKD, however, we fight against a possible active adversary.



Reduce channel loss via a “fair-sampling device” (leaves only problem of detection efficiency)

N. Gisin, S. Pironio and N. Sangouard, Phys. Rev. Lett. 105, 070501 (2010); N. Sangouard et al., Phys. Rev. Lett. 106, 120403 (2011); M. Curty and T. Moroder, Phys. Rev. A 84, 010304(R) (2011).

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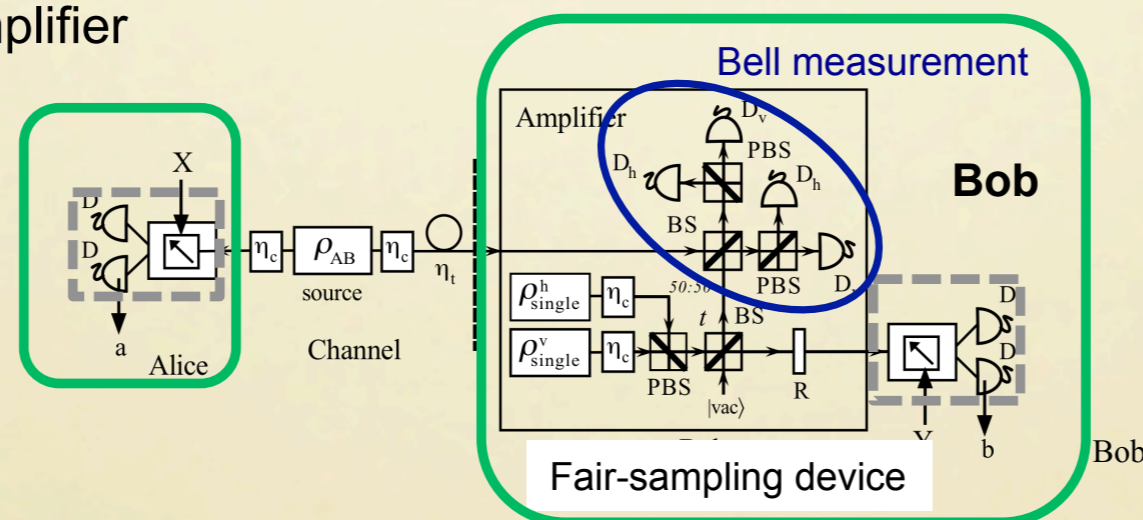
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Reduce channel loss via a “fair-sampling device” (leaves only problem of detection efficiency)

Heralded qubit amplifier

For simplicity, qubit amplifier only on Bob's side



A simpler quantum relay works as well even with SPDC sources!

N. Gisin, S. Pironio and N. Sangouard, Phys. Rev. Lett. 105, 070501 (2010); N. Sangouard et al., Phys. Rev. Lett. 106, 120403 (2011); M. Curty and T. Moroder, Phys. Rev. A 84, 010304(R) (2011).

SIDE-CHANNELS

What performance can we expect in practice?

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What performance can we expect in practice?

Simulation with

- Full-mode analysis [in contrast to perturbation approach]
- Detector, coupling efficiencies
- Optimization over variable parameters

Equipment:

- * Standard PDC as entangled & heralded PDC as single photon sources
- * Photon number resolving detectors

M. Curty and T. Moroder, Phys. Rev. A 84, 010304(R) (2011).

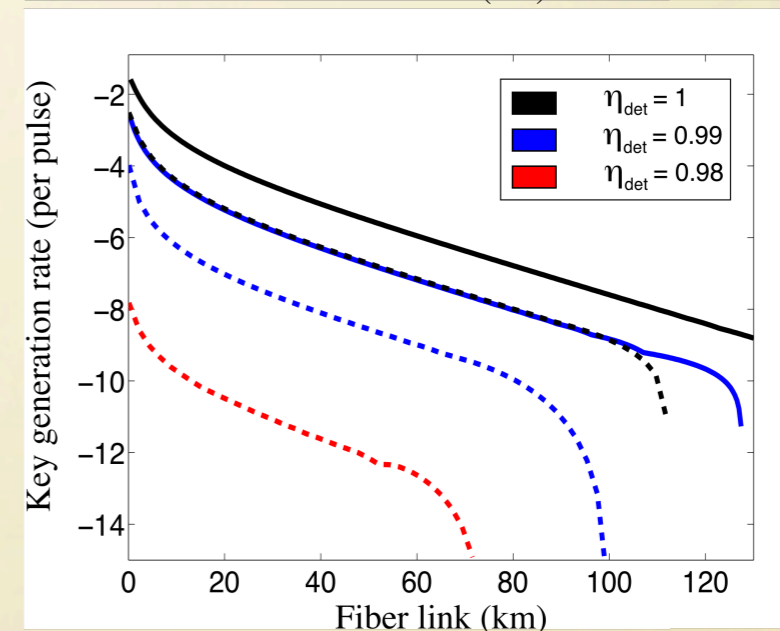
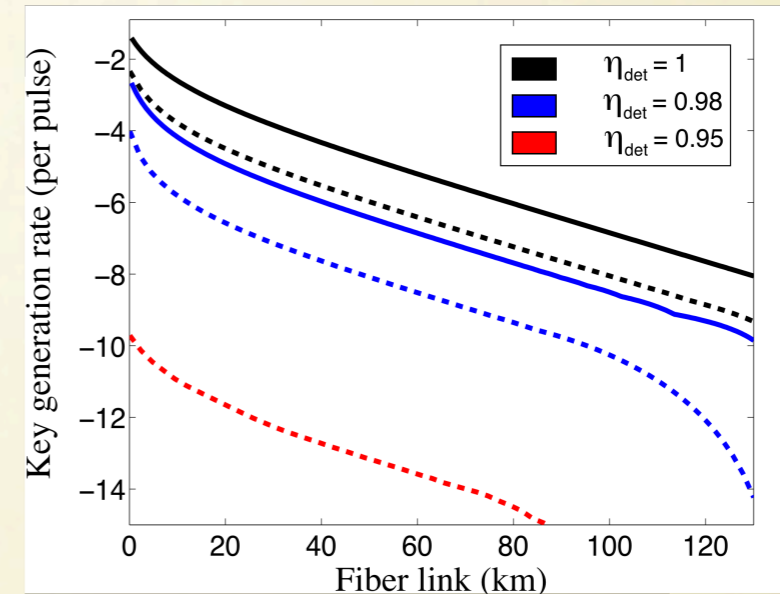
See also: D. Pitkänen et al., Phys. Rev. A 84, 022325 (2011).

Limitations:

Requires near unity detection efficiency

An extremely low key rate (of order 10^{-8} - 10^{-10} per pulse) at practical distances

di-QKD is a very beautiful idea but **impractical** with current technology => Need to improve entanglement sources, couplers and detectors!



“Original” qubit amplifier (dashed line) quantum relay (solid line). Upper figure shows a security analysis from Gisin et al. [PRL **105**, 070501 (2010)]. Lower figure shows the conservative situation of assigning inconclusive to conclusive results deterministically.

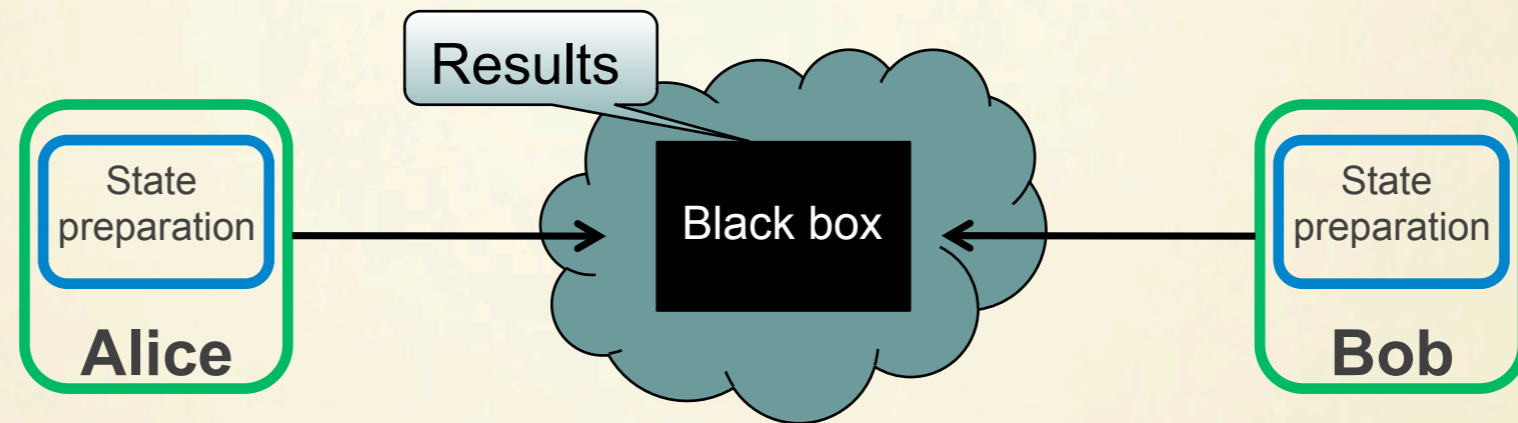
SIDE-CHANNELS

Rethink the problem: Most side channel attacks occur in the detectors

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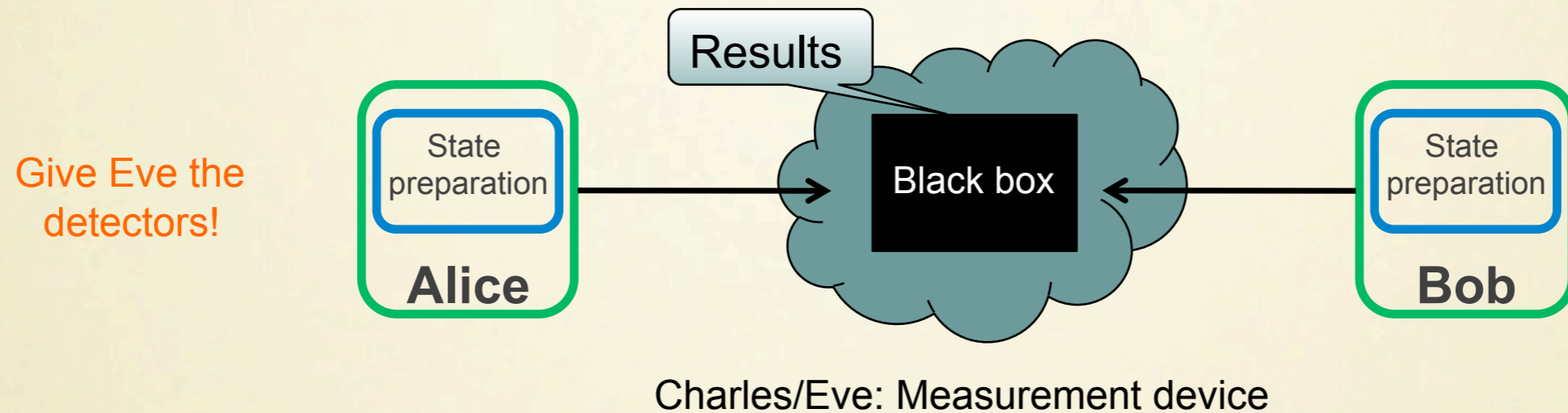
Give Eve the detectors!



Charles/Eve: Measurement device

SIDE-CHANNELS

Rethink the problem: Most side channel attacks occur in the detectors



Measurement-device independent QKD

A practical way to do QKD with “untrusted detectors”

Automatically immune to all detector side-channel attacks (existing and yet to be discovered)

No need to certify the measurement device (it can be even manufactured by a malicious eavesdropper, Eve). This is good news for QKD standardisation and certification by European Telecommunications Standards Institute (ETSI)

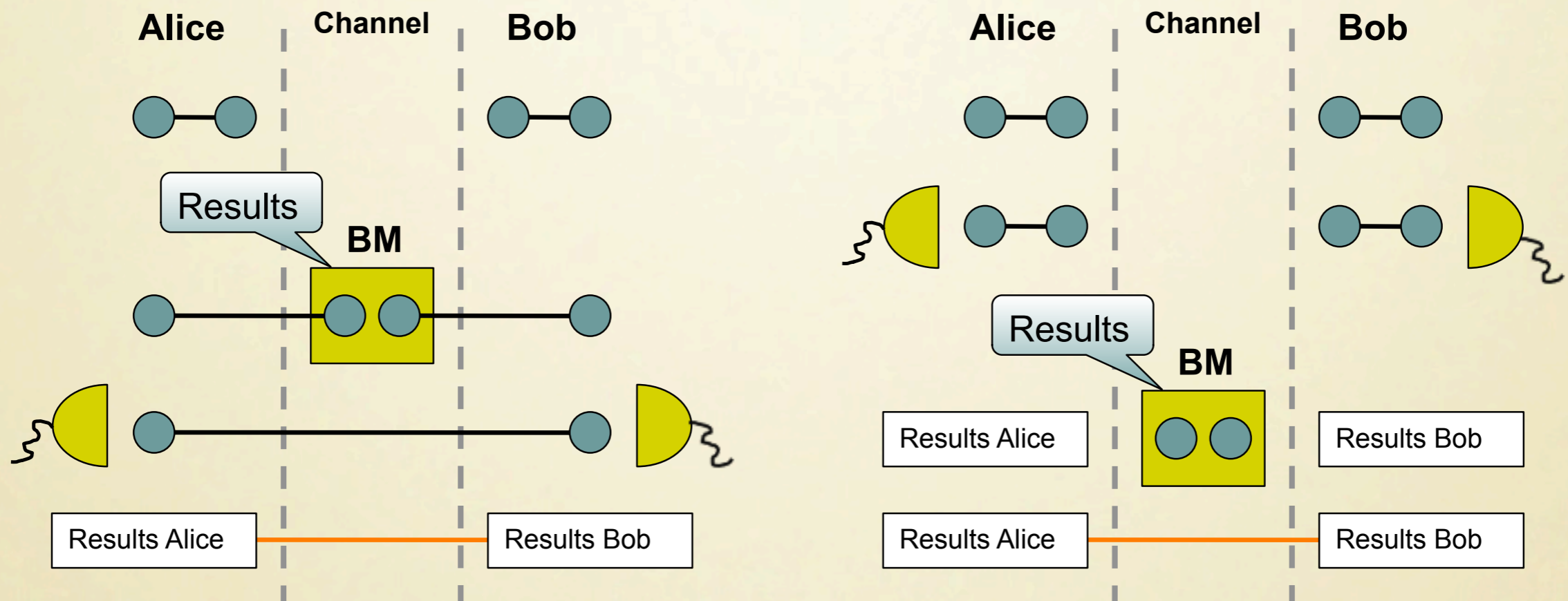
H.-K. Lo, M. Curty and B. Qi, Phys. Rev. Lett. 108, 130503 (2012); E. Biham, B. Huttner and T. Mor, Phys. Rev. A 54, 2651-2658 (1996); H. Inamori, Algorithmica 34, 340-365 (2002).

SIDE-CHANNELS

Intuition why it can be secure:

SIDE-CHANNELS

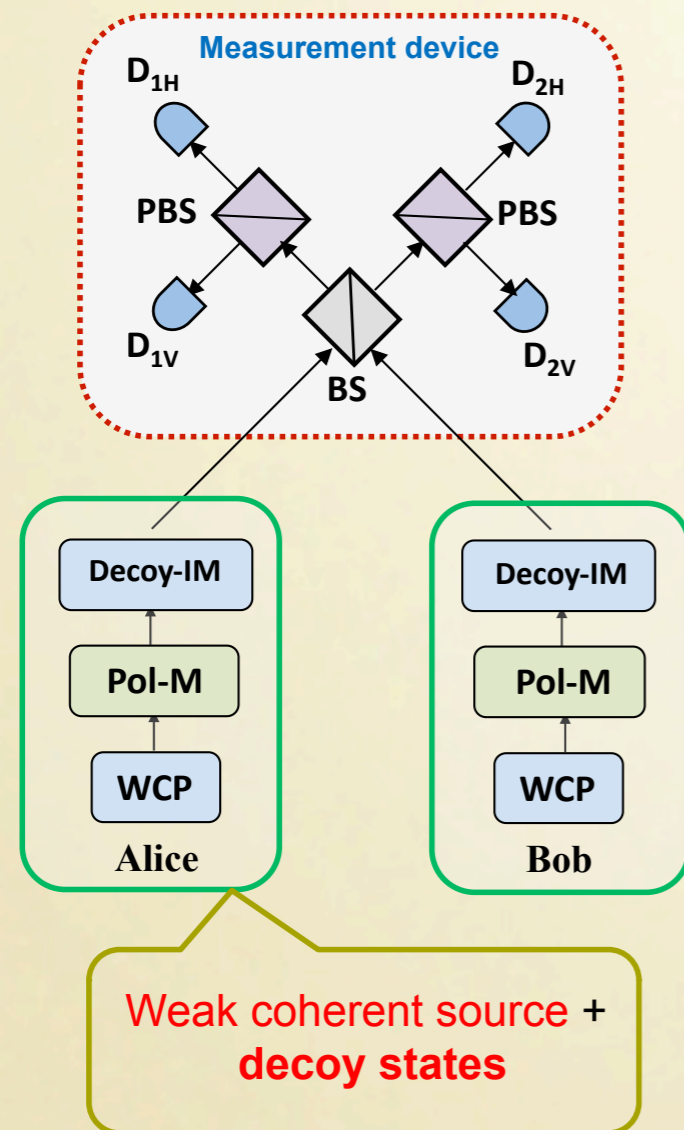
Intuition why it can be secure:



The result of the Bell measurement reveals correlations between Alice and Bob's bits but not the value of the individual bits

SIDE-CHANNELS

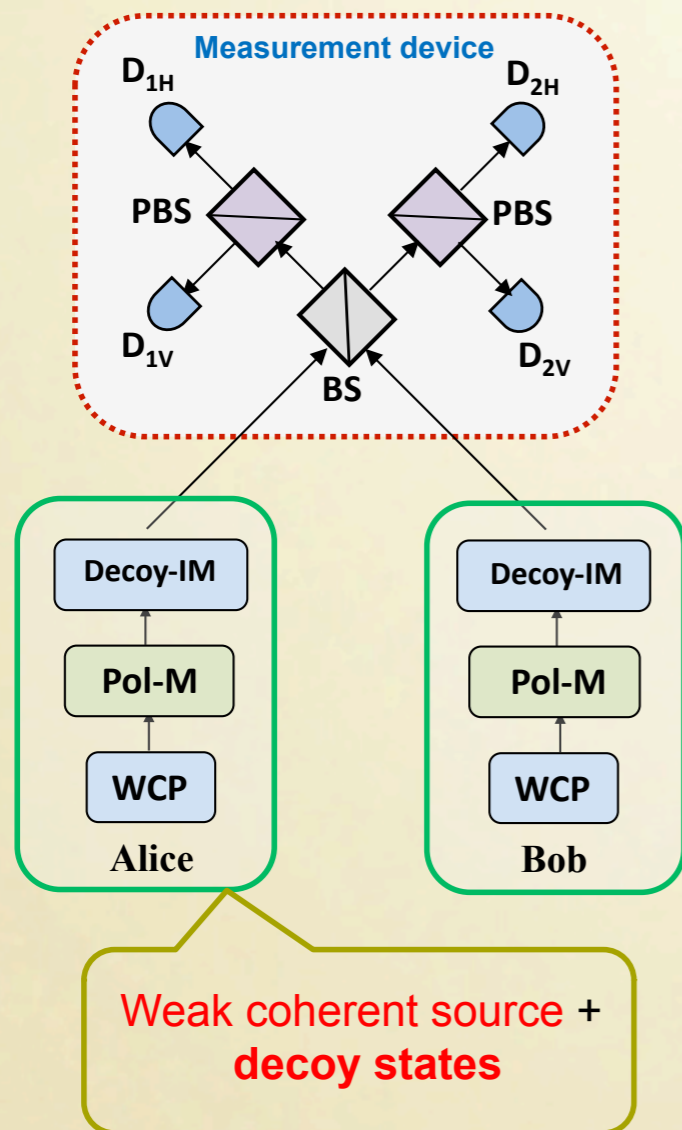
Measurement-device independent QKD



H.-K. Lo, M. Curty and B. Qi, Phys. Rev. Lett. 108, 130503 (2012).

SIDE-CHANNELS

Measurement-device independent QKD



$$R \geq p_{1,1,Z} Y_{1,1,Z} [1 - h(e_{1,1,X})] - Q_Z h(E_Z)$$

Z basis for key generation

X basis for testing only

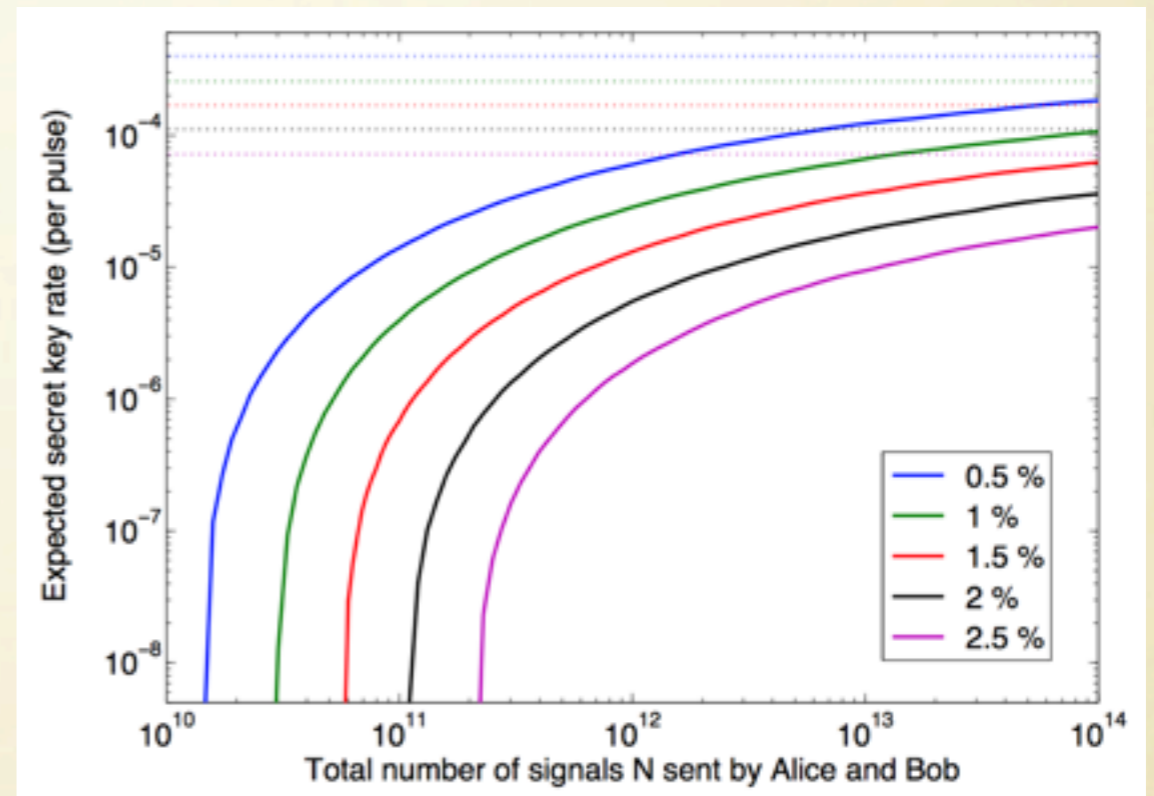
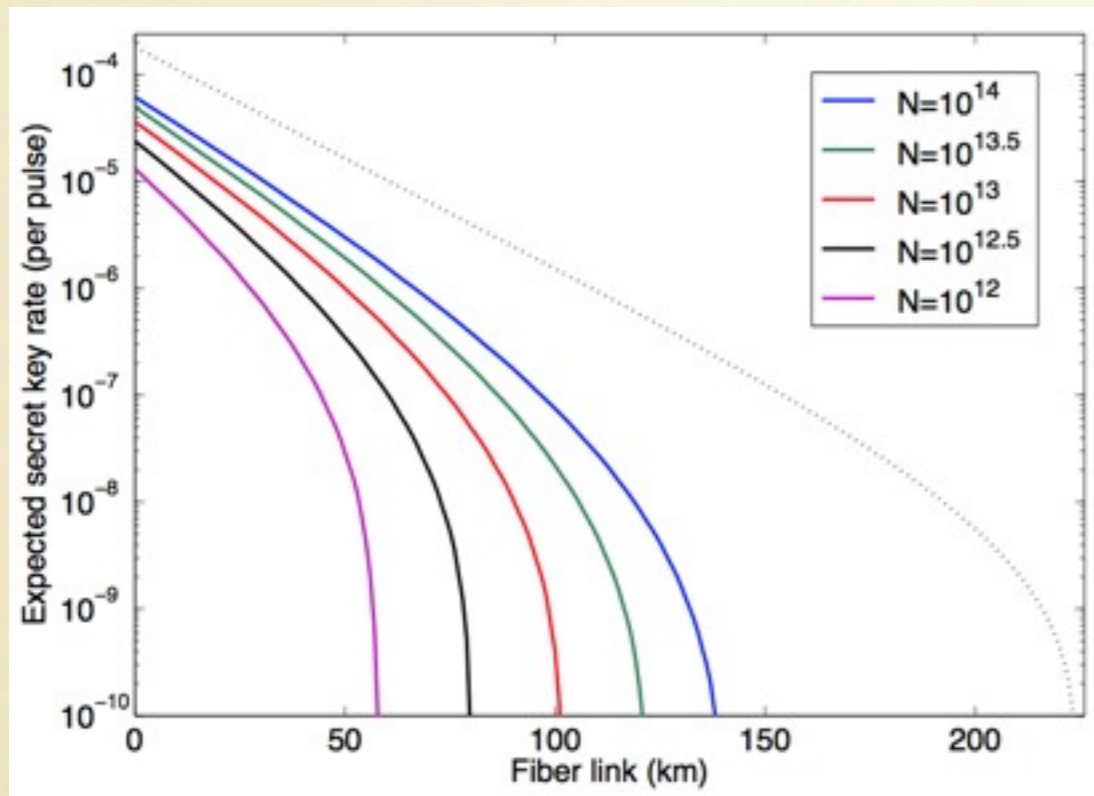
Q_Z and E_Z can be measured directly from the experiment.

$Y_{1,1,Z}$ and $e_{1,1,X}$ are estimated using decoy states

H.-K. Lo, M. Curty and B. Qi, Phys. Rev. Lett. 108, 130503 (2012).

SIDE-CHANNELS

Simulation results (finite-key case):



The experimental parameters are: $\alpha = 0.2$ dB/km, $\eta_B = 14.5\%$, $Y_0 = 6.02 \times 10^{-6}$ and the security bound $\epsilon = 10^{-10}$. The misalignment in the first figure is 1.5%

If Alice and Bob use laser diodes at 1 GHz repetition rate, and each of them sends $N = 10^{13}$ signals, we find, for instance, that they can distribute a 1 Mb secret key over a 75 km fiber link in less than 3 hours.

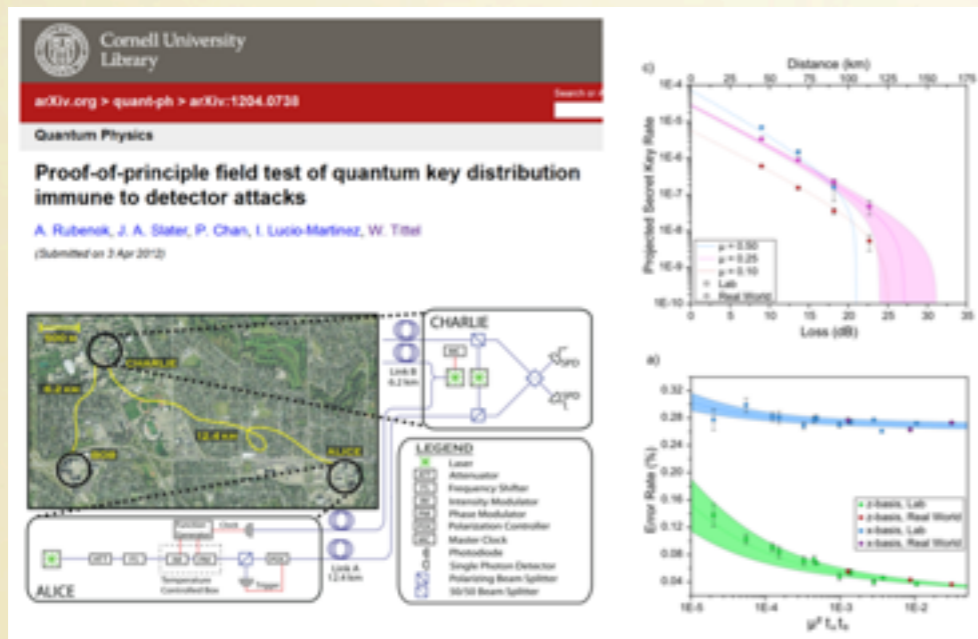
M. Curty et al., preprint arXiv:1307:1081.

SIDE-CHANNELS

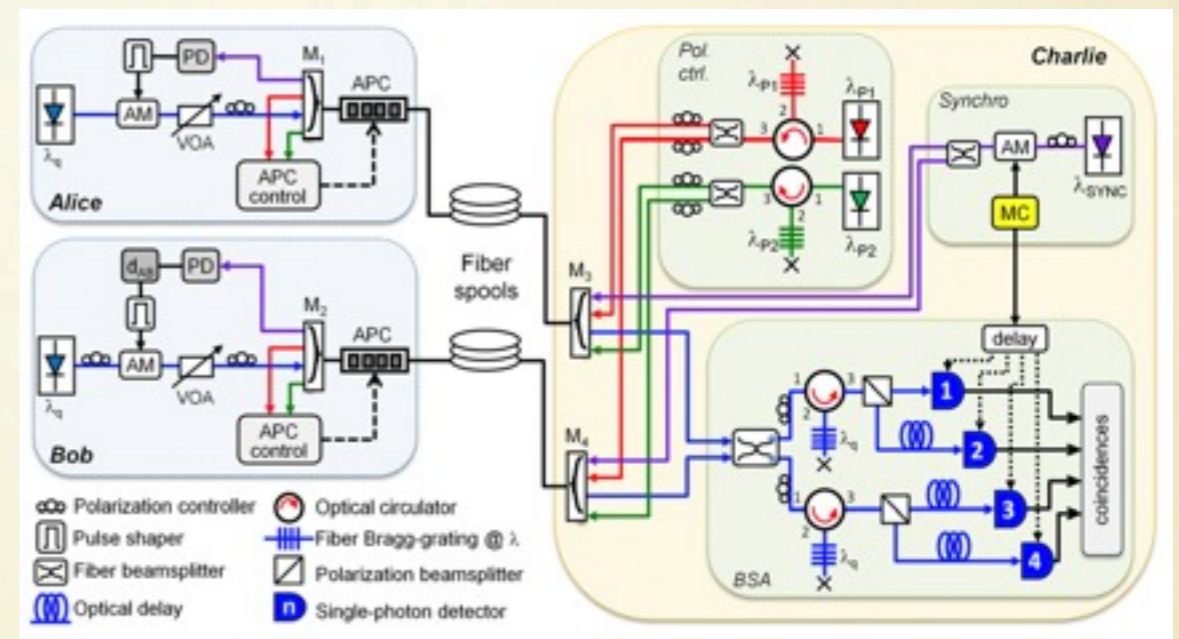
Let's return to the lab...

SIDE-CHANNELS

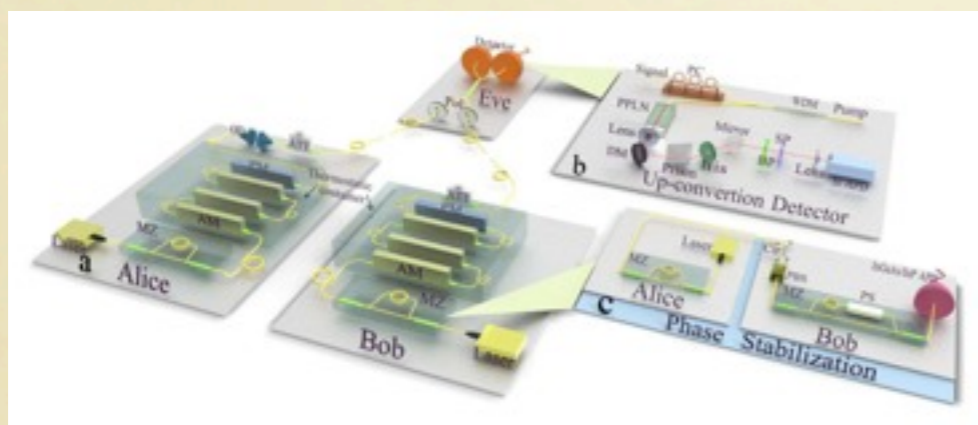
Let's return to the lab...



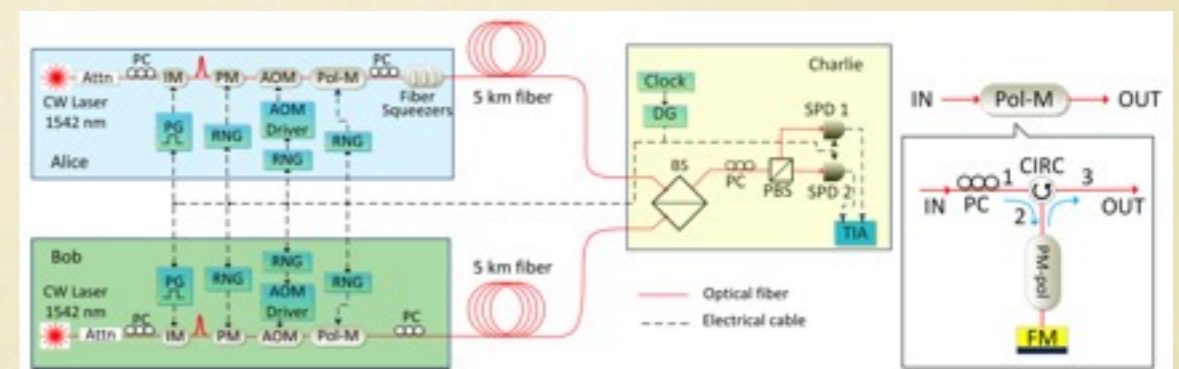
A. Rubenok et al., preprint arXiv:1204.0738



T. Ferreira da Silva et al., preprint arXiv:1207.6345



Y. Liu et al., preprint arXiv:1209.6178



Z. Tang et al., preprint arXiv:1306.6134



THANK YOU FOR YOUR
ATTENTION